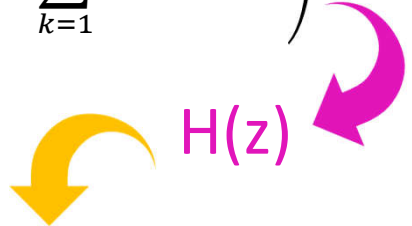


Filtros digitais

# **DESIGN DE FILTROS DIGITAIS**

# Introdução

- Filtros digitais são uma classe particular de sistemas LTI
- Filtro seletor de frequência sugere algo que seleciona uma faixa de frequência e rejeita outras, mas filtros são mais que isso, **qualquer sistema que modifique frequências relativamente a outras é chamado filtro**

$$y[n] = \frac{1}{a_0} \left( \sum_{k=0}^M b_k x[n-k] - \sum_{k=1}^N a_k y[n-k] \right)$$


## IIR

- $H(z)$  aproximada por uma razão de polinômios

## FIR

- $H(z)$  aproximada por um polinômio

# Introdução

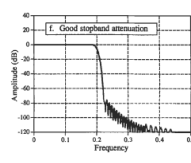
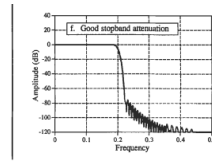
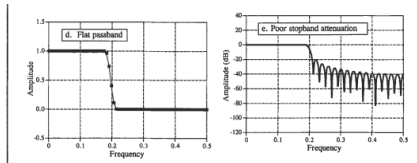
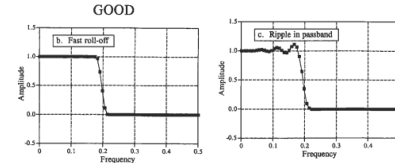
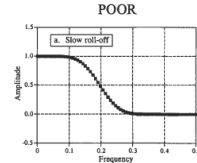
- Filtros digitais são usados com duas finalidades principais
  - Separar sinais que foram combinados
  - Restaurar sinais que foram distorcidos de alguma maneira

# Introdução

- São muito importantes em PDS, sua extraordinária performance tornou o PDS tão popular
- Por que utilizar filtros digitais ao invés de analógicos?
  - Filtros analógicos são baratos e tem uma ampla faixa e atuação em amplitude e frequência
  - **Filtros digitais tem performances milhares de vezes melhores que um analógico!**

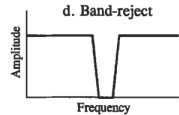
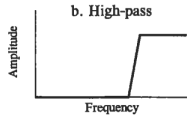
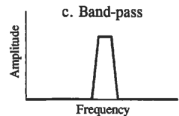
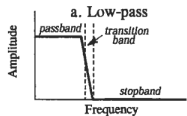
# Como filtrar um sinal?

- Depende da aplicação
- Características dos sinais de entrada e saída



Especificação das propriedades do sistema

- Forma, frequência de corte com base no filtro ideal



Gerar o núcleo do filtro

- Aproximação das propriedades por meio de uma sequência causal de tempo discreto que aproxime a RF desejada

$$y[n] = \frac{1}{a_0} \left( \sum_{k=0}^M b_k x[n-k] - \sum_{k=1}^N a_k y[n-k] \right)$$

$h[n]$

Avaliar a RF do filtro gerado em 2

- $H(z)$  aproxima corretamente o filtro ideal
- Parâmetros que determinam uma boa  $H(z)$ :
  - Roll-off
  - Ripple
  - Atenuação

Filtrar o sinal

- Depende da tecnologia

$H(z)$

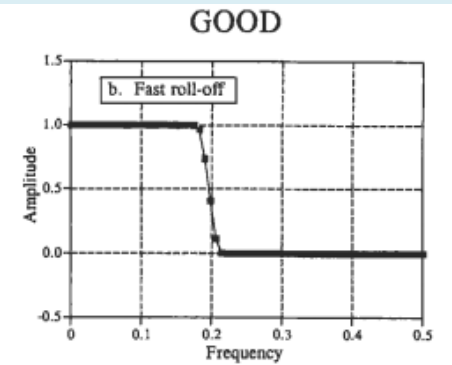
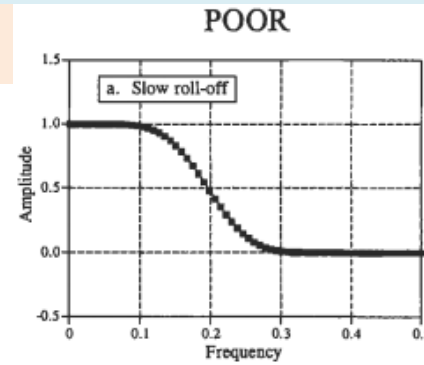
$h[n]$

# Parâmetros para o domínio da frequência

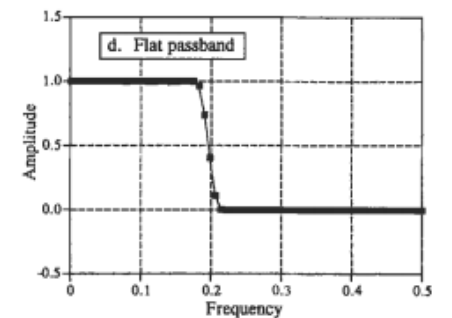
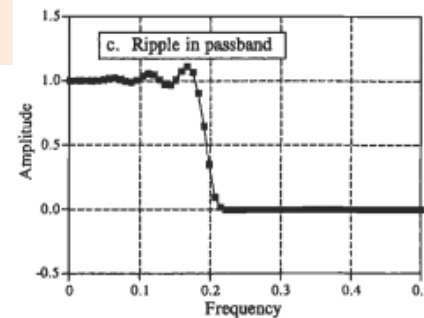
- **Roll-off rápido**: para separar duas frequências próximas
- Não pode haver ondulação (**ripple**) no passband
- **Boa atenuação** no stopband
- Por que não há nada sobre a fase?
  - A fase não é tão importante em aplicações no domínio da frequência

# Parâmetros para o domínio da frequência

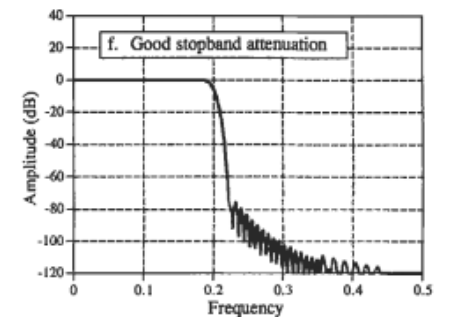
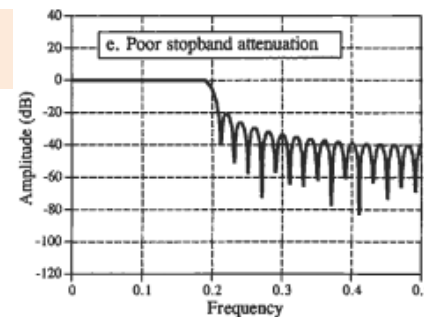
Roll-off



ripple



Atenuação



# Parâmetros para o domínio da frequência

- A proposta destes filtros é permitir que **algumas frequências passem inalteradas e outras sejam eliminadas**
  - *Passband*: frequências que passam
  - *Stopband*: frequências que são bloqueadas
  - *Banda de transição*: entre as 2 anteriores
    - *Roll-off rápido*: banda de transição pequena
    - Frequência de corte (*cutoff*): divisão entre a banda de passagem e a de transição



# Design de filtros digitais

## Resposta Impulsiva (núcleo do filtro)

- A maneira mais simples de **implementar filtros** digitais é **convoluir** o **signal de entrada** com a **resposta impulsiva do filtro**

– Quando a RI é usada com esta finalidade é chamada de **núcleo do filtro**

**FIR**

- Em um **filtro implementado por convolução**, cada amostra da saída é calculada “pesando-se” as amostras de entradas e adicionando-as

# Design de filtros digitais

## Recursão

- Filtros recursivos são uma extensão dos de convolução, só que eles **utilizam informação das saídas anteriores** além de pontos da entrada, ao invés de utilizarem o núcleo, utilizam uma serie de coeficientes recursivos

- É importante salientar que **TODOS os filtros tem uma RI, mesmo que ela não tenha sido usada para implementar o filtro**

– Para conhecer a RI de um filtro recursivo, alimente-o com um sinal de entrada e observe a saída

# Design de filtros digitais

## Recursão

- A RI de filtros recursivos é composta por senóides que decaem exponencialmente em amplitude

IIR

- Isso faz com que tenham RI infinitamente longa

# IIR x FIR

## Recursão

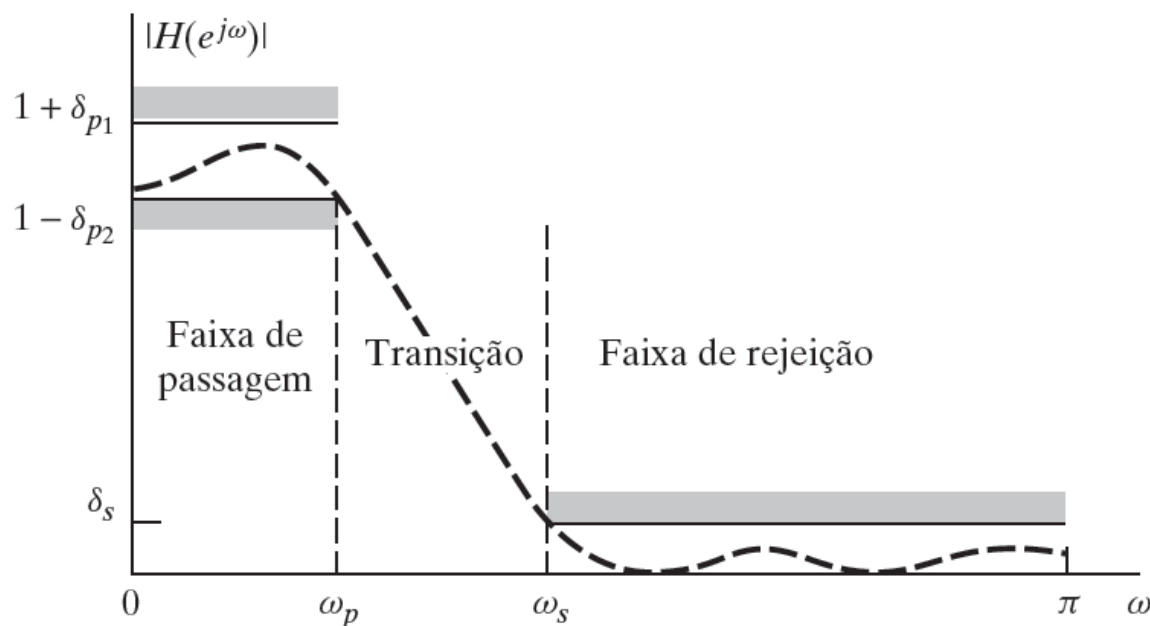
- Por causa dessa característica, filtros recursivos são chamados de Filtros de Resposta Impulsiva Infinita – **IIR** (*Infinite Impulse Response*)

## Convolução

- Por outro lado, os filtros obtidos por convolução são chamados de Filtros de Resposta Impulsiva Finita – **FIR** (*Finite Impulse Response*)

## Especificações do filtro

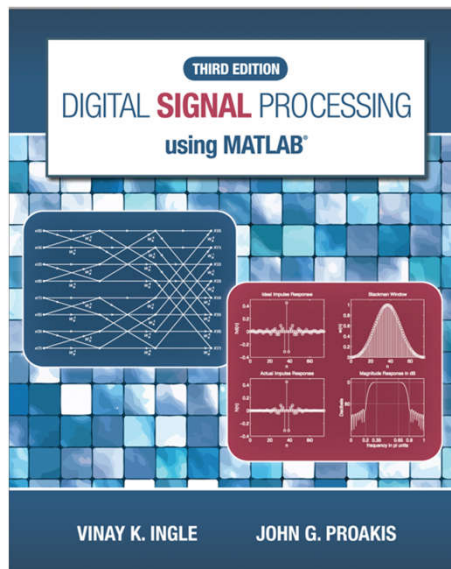
- Diagrama de tolerâncias de um filtro passa-baixas.



# Design de filtros

## MATLAB

- Capitulo 7 – FIR
- Capitulo 8 - IIR



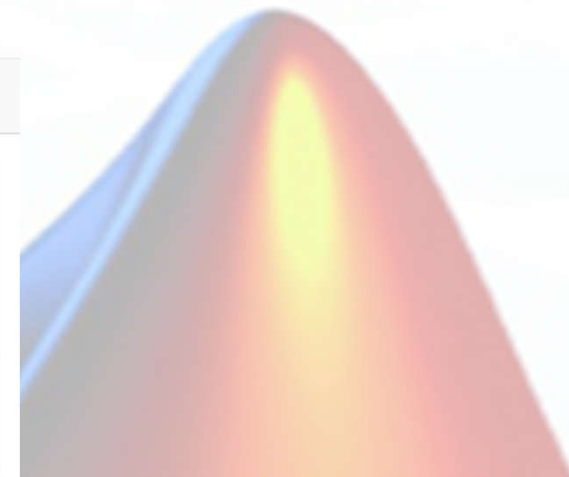
## Teoria

- Capitulo 7



▼ IIR Filters

<a href="#">butter</a>	Butterworth filter design
<a href="#">buttord</a>	Butterworth filter order and cutoff frequency
<a href="#">cheby1</a>	Chebyshev Type I filter design
<a href="#">cheb1ord</a>	Chebyshev Type I filter order
<a href="#">cheby2</a>	Chebyshev Type II filter design
<a href="#">cheb2ord</a>	Chebyshev Type II filter order
<a href="#">designfilt</a>	Design digital filters
<a href="#">ellip</a>	Elliptic filter design
<a href="#">ellipord</a>	Minimum order for elliptic filters
<a href="#">polyscale</a>	Scale roots of polynomial
<a href="#">polystab</a>	Stabilize polynomial
<a href="#">yulewalk</a>	Recursive digital filter design



▼ FIR Filters

<a href="#">cfirpm</a>	Complex and nonlinear-phase equiripple FIR filter design
<a href="#">designfilt</a>	Design digital filters
<a href="#">fir1</a>	Window-based FIR filter design
<a href="#">fir2</a>	Frequency sampling-based FIR filter design
<a href="#">fircls</a>	Constrained-least-squares FIR multiband filter design
<a href="#">fircls1</a>	Constrained-least-squares linear-phase FIR lowpass and highpass filter design
<a href="#">firls</a>	Least-squares linear-phase FIR filter design
<a href="#">firpm</a>	Parks-McClellan optimal FIR filter design
<a href="#">firpmord</a>	Parks-McClellan optimal FIR filter order estimation
<a href="#">gaussdesign</a>	Gaussian FIR pulse-shaping filter design
<a href="#">intfilt</a>	Interpolation FIR filter design
<a href="#">kaiserord</a>	Kaiser window FIR filter design estimation parameters
<a href="#">maxflat</a>	Generalized digital Butterworth filter design
<a href="#">rcosdesign</a>	Raised cosine FIR pulse-shaping filter design
<a href="#">sgolay</a>	Savitzky-Golay filter design

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Filtros digitais

# **DESIGN DE FILTROS DIGITAIS**

## **EXEMPLOS FIR**



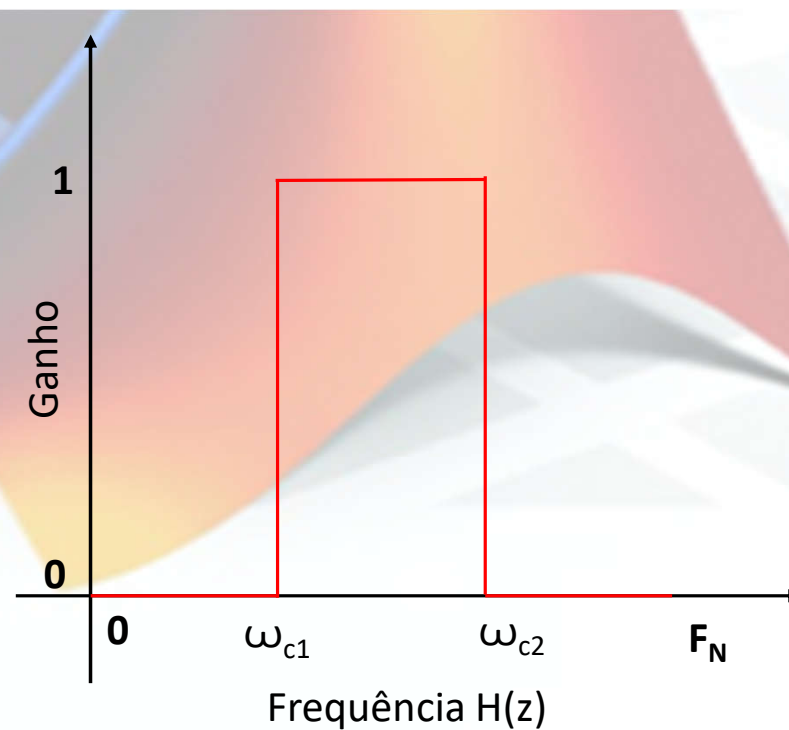
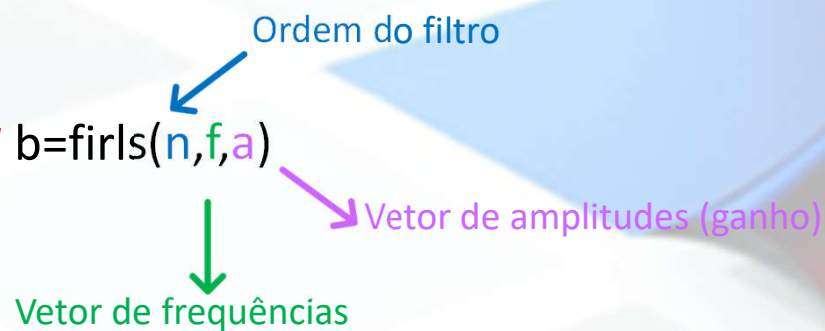
# firls

## Description

`b = firls(n,f,a)` returns row vector `b` containing the  $n+1$  coefficients of an order- $n$  FIR filter. The frequency and amplitude characteristics of the resulting filter match those given by vectors `f` and `a`. [example](#)

`b = firls(n,f,a,w)` uses `w` to weigh the frequency bins. [example](#)

`b = firls(__,ftype)` designs antisymmetric (odd) filters, where `ftype` specifies the filter as a differentiator or Hilbert transformer. You can use `ftype` with any of the previous input syntaxes. [example](#)



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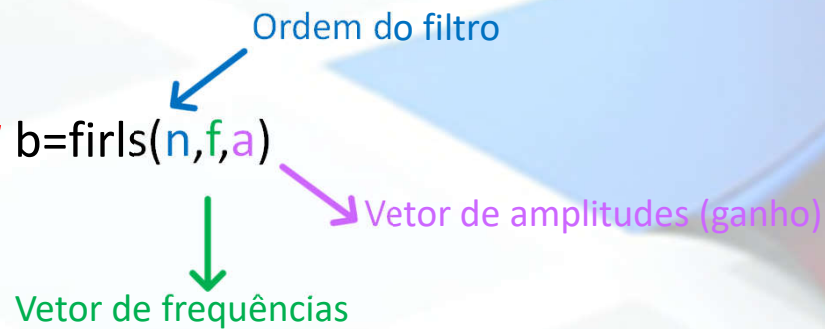
# firls

## Description

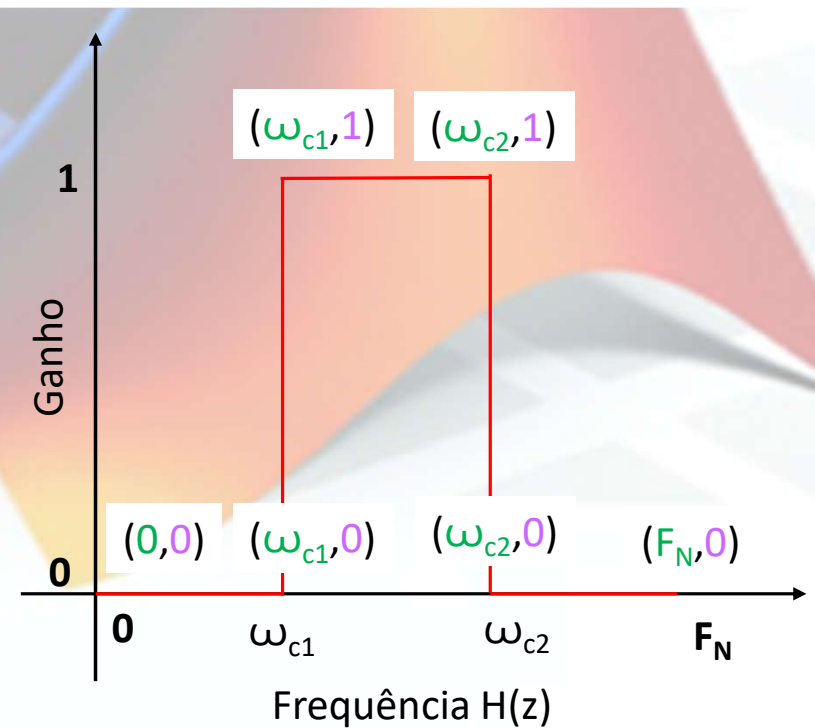
`b = firls(n,f,a)` returns row vector `b` containing the  $n+1$  coefficients of an order- $n$  FIR filter. The frequency and amplitude characteristics of the resulting filter match those given by vectors `f` and `a`. [example](#)

`b = firls(n,f,a,w)` uses `w` to weigh the frequency bins. [example](#)

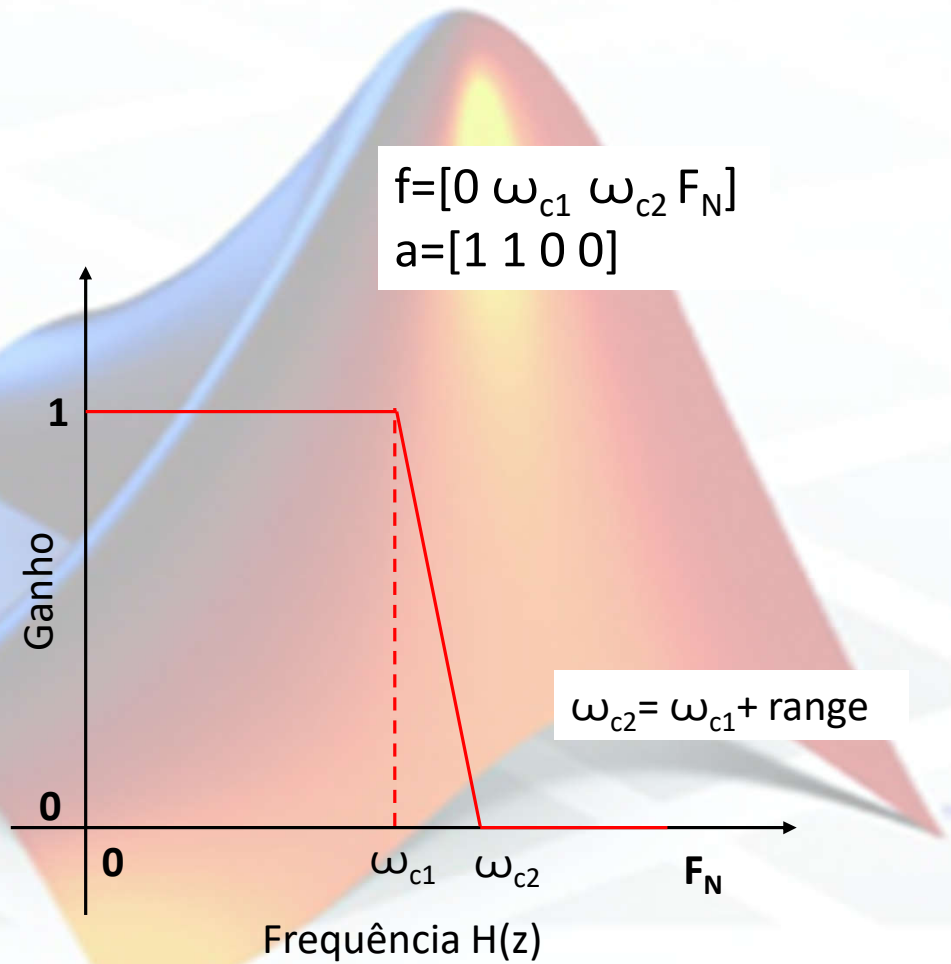
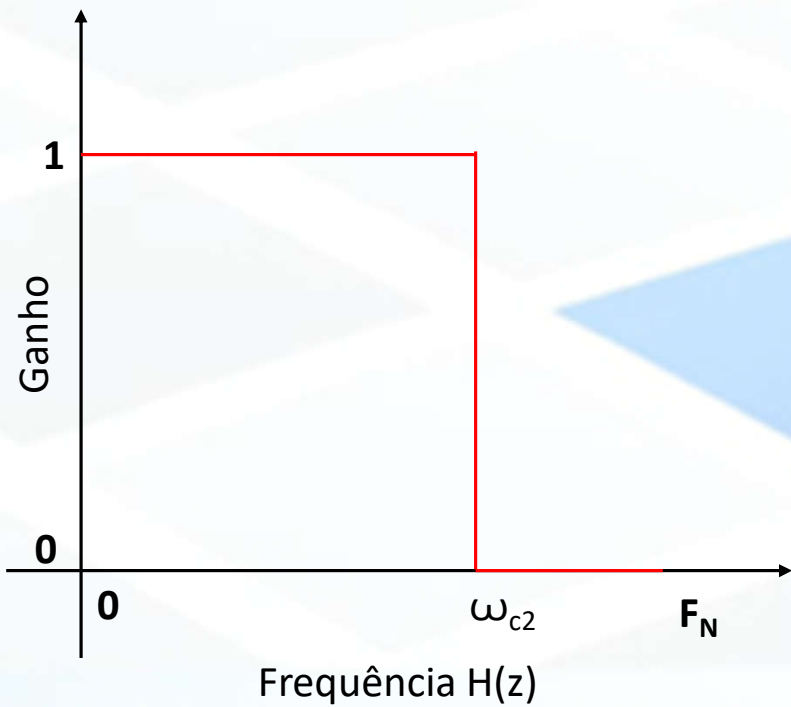
`b = firls(__,ftype)` designs antisymmetric (odd) filters, where `ftype` specifies the filter as a differentiator or Hilbert transformer. You can use `ftype` with any of the previous input syntaxes. [example](#)



$f=[0 \ \omega_{c1} \ \omega_{c1} \ \omega_{c2} \ \omega_{c2} \ F_N]$   
 $a=[0 \ 0 \ 1 \ 1 \ 0 \ 0]$



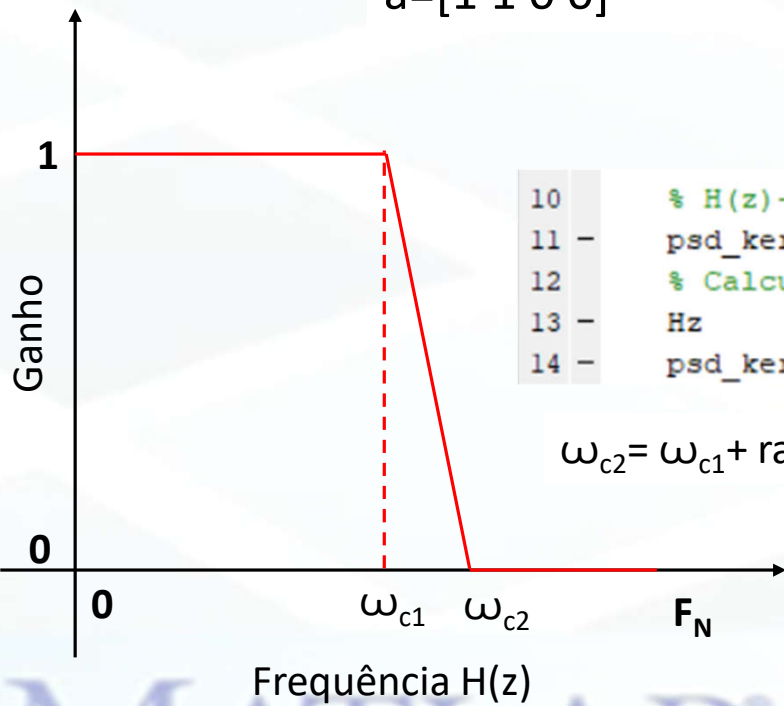
# Filtro ideal



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# firls

$$f=[0 \ \omega_{c1} \ \omega_{c2} \ F_N]$$
$$a=[1 \ 1 \ 0 \ 0]$$



```
1 - clear
2 - Fs = 1024; % Hz
3 - nyquist = Fs/2;
4 - f = [0 0.25 0.3 1];
5 - a = [1 1 0 0];
6 - ordem=50;
7 - % filter kernel
8 - kern = firls(ordem,f,a);
```

```
10 % H(z)- PSD do kernel
11 - psd_kern = abs(fft(kern)).^2;
12 % Calcula o vetor de frequencias e exhibe apenas a parte positiva (simetria)
13 - Hz = linspace(0,Fs/2,floor(length(kern)/2)+1);
14 - psd_kern = psd_kern(1:length(Hz));
```

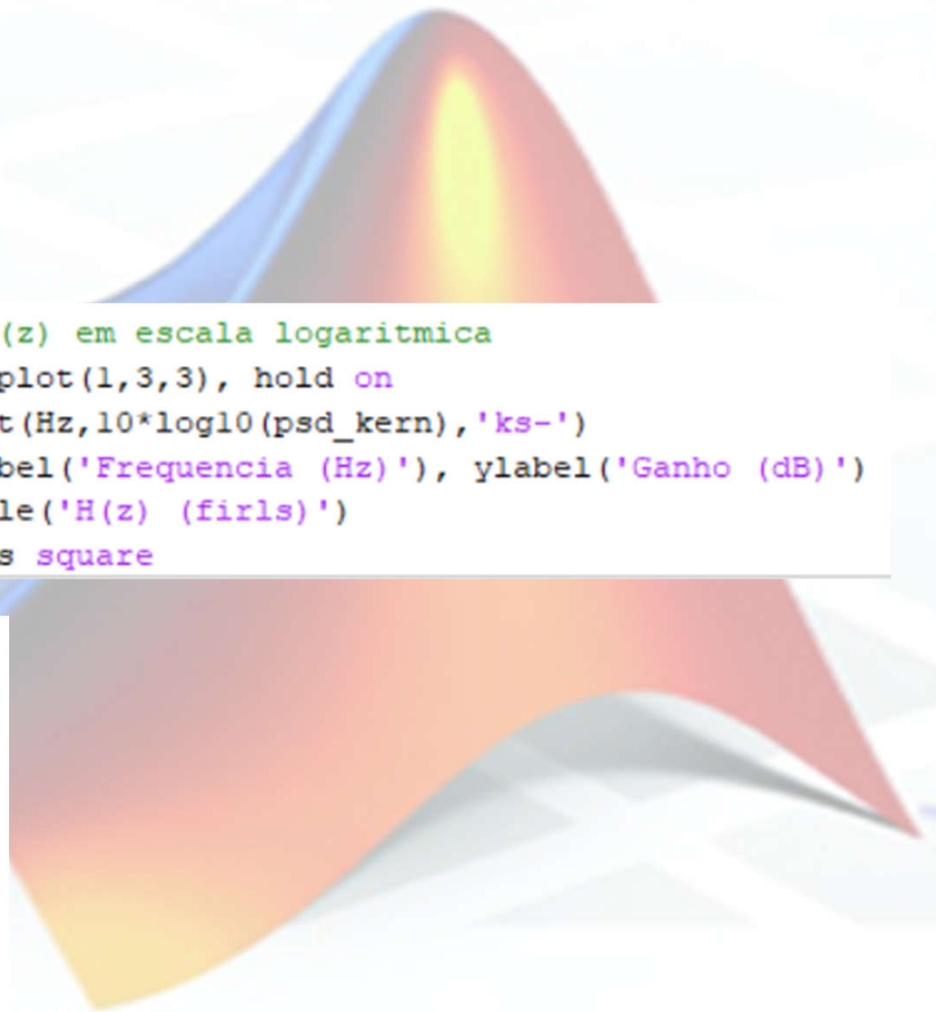
MATLAB®

# firls

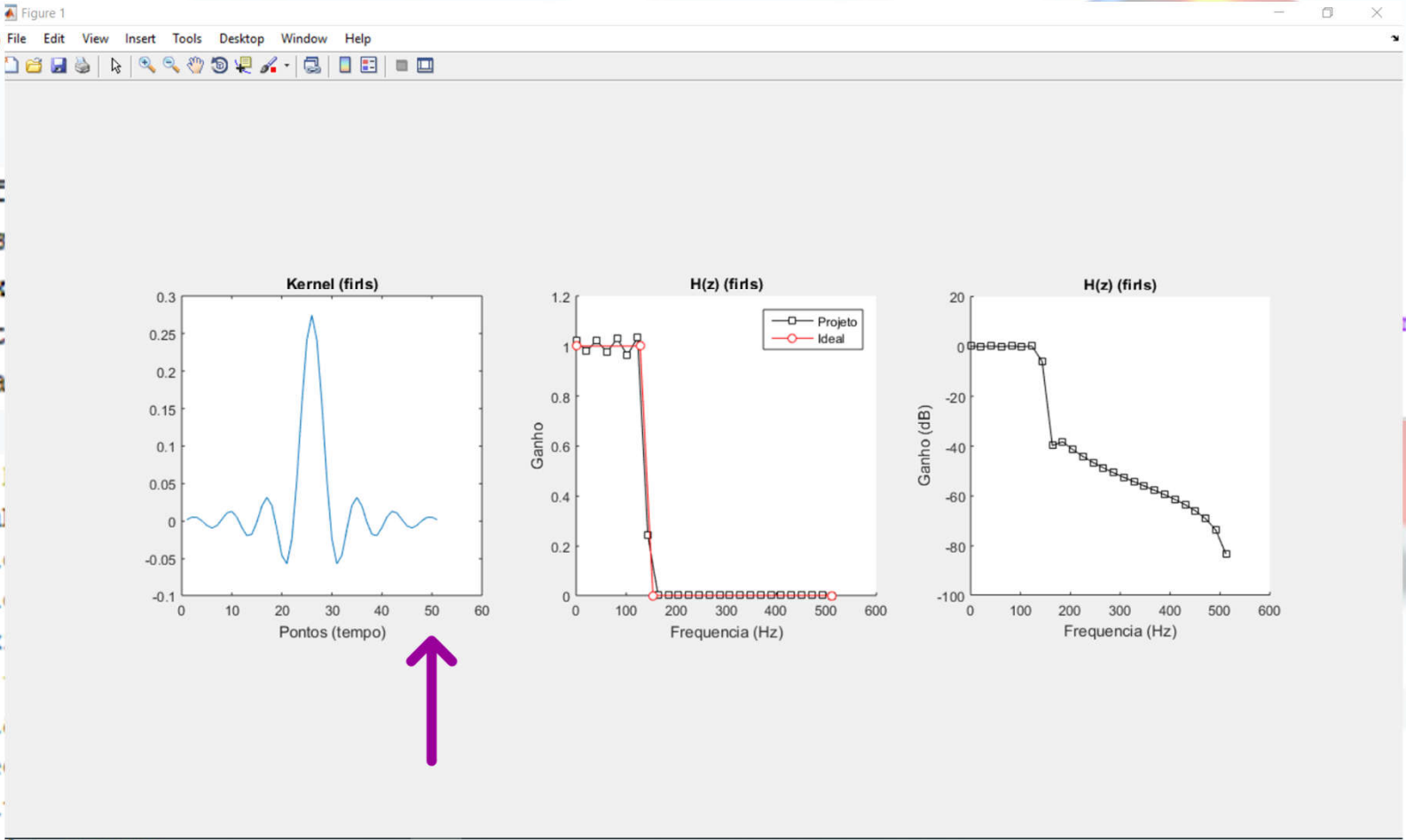
```
16 - figure(1);
17 - subplot(1,3,1);plot(kern)
18 - xlabel('Pontos (tempo)')
19 - title('Kernel (firls)')
20 - axis square
```

```
22 - % H(z) (PSD)
23 - subplot(1,3,2), hold on
24 - plot(Hz,psd_kern,'ks-','markerfacecolor','w')
25 - plot(f*nyquist,a,'ro-','markerfacecolor','w')
26 - axis square
27 - % titulo, legenda
28 - xlabel('Frequencia (Hz)'), ylabel('Ganho')
29 - legend({'Projeto';'Ideal'})
30 - title('H(z) (firls)')
31 - axis square
```

```
33 - % H(z) em escala logaritmica
34 - subplot(1,3,3), hold on
35 - plot(Hz,10*log10(psd_kern),'ks-')
36 - xlabel('Frequencia (Hz)'), ylabel('Ganho (dB)')
37 - title('H(z) (firls)')
38 - axis square
```



firls



```
16 - f
17 - s
18 - x
19 - t
20 - a

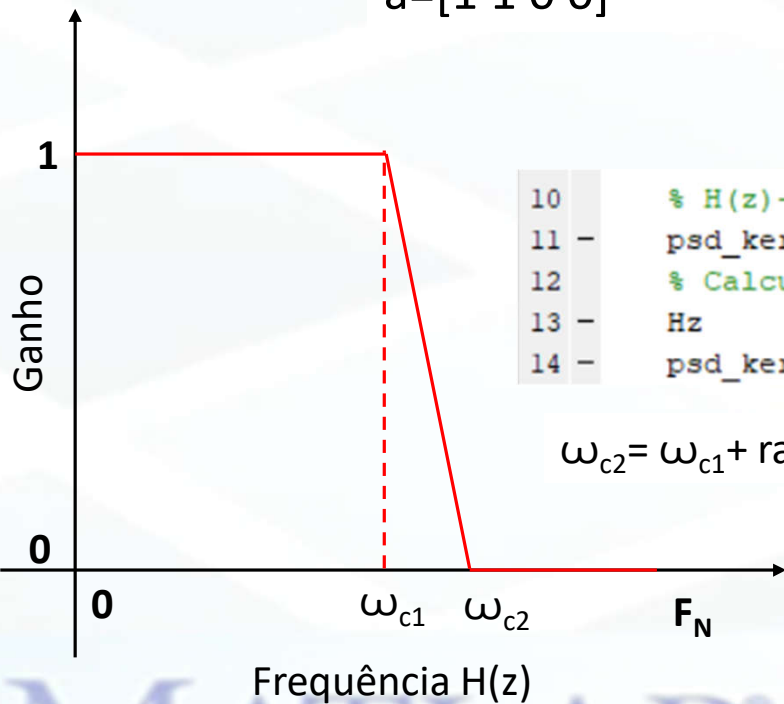
22 - %
23 - su
24 - pl
25 - pl
26 - ax
27 - %
28 - xl
29 - le
30 - ti
31 - axis square
```

anho (dB)')



# Filtro ideal

$$f=[0 \ \omega_{c1} \ \omega_{c2} \ F_N]$$
$$a=[1 \ 1 \ 0 \ 0]$$

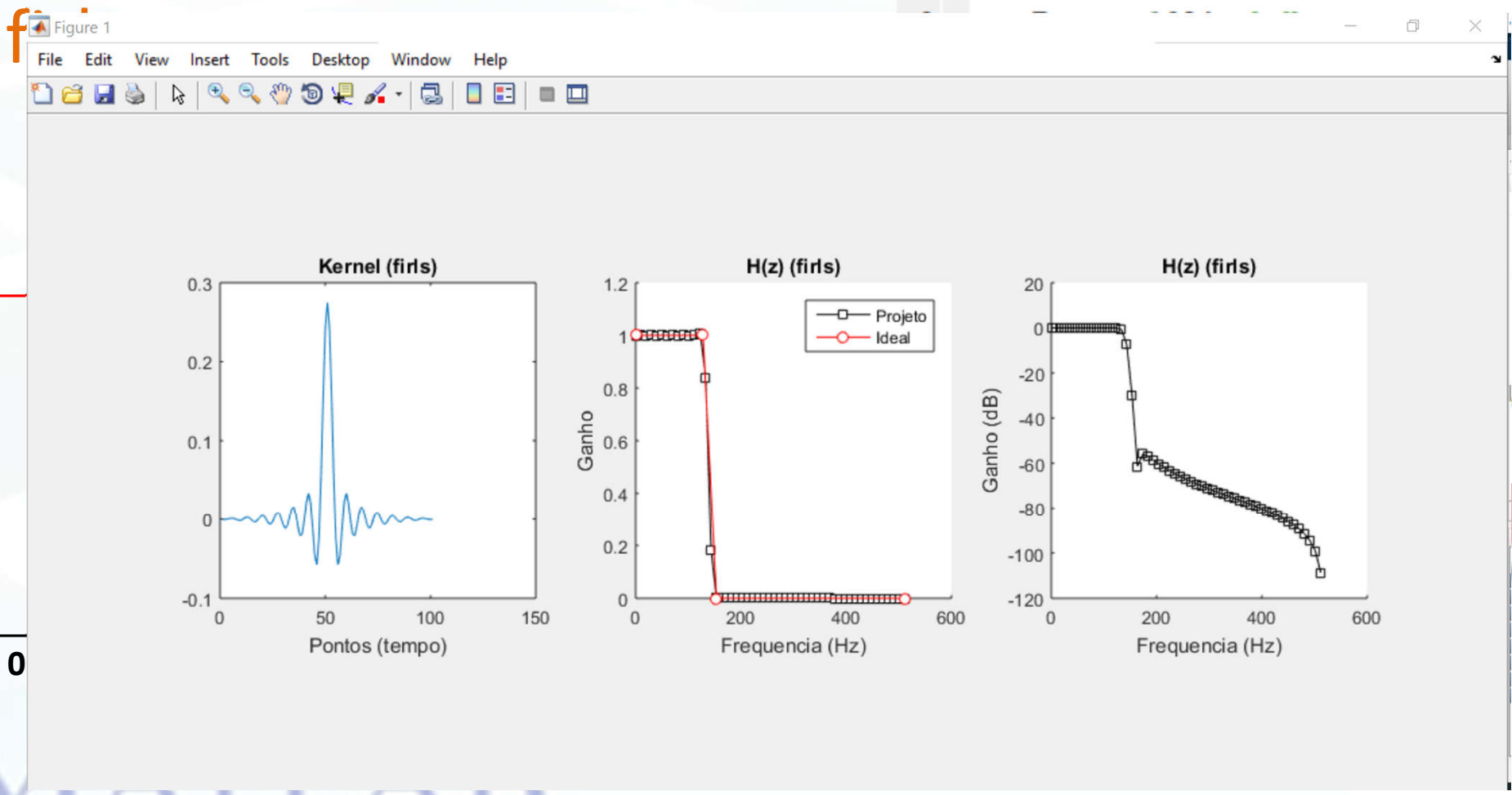


```
1 - clear
2 - Fs = 1024; % Hz
3 - nyquist = Fs/2;
4 - f = [0 0.25 0.3 1];
5 - a = [1 1 0 0];
6 - ordem=50; ←
7 - % filter kernel
8 - kern = firls(ordem,f,a);
```

```
10 % H(z)- PSD do kernel
11 - psd_kern = abs(fft(kern)).^2;
12 % Calcula o vetor de frequencias e exhibe apenas a parte positiva (simetria)
13 - Hz = linspace(0,Fs/2,floor(length(kern)/2)+1);
14 - psd_kern = psd_kern(1:length(Hz));
```

MATLAB®

```
1 - clear
```



Ganho  
1  
0  
0

etria)



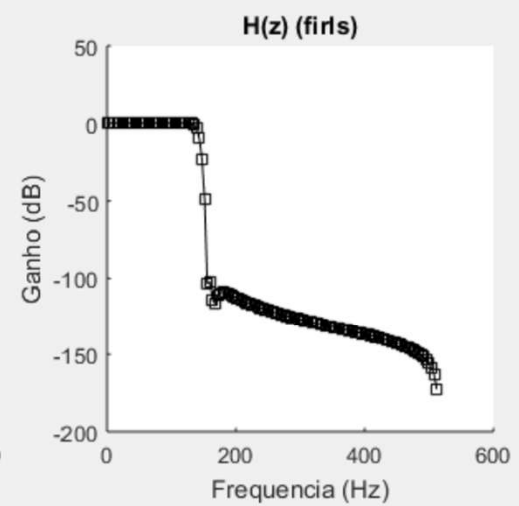
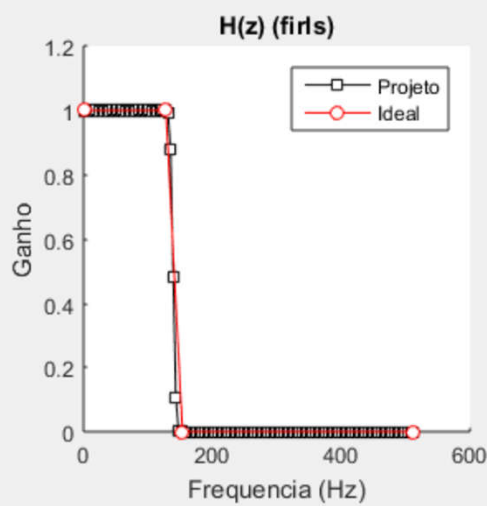
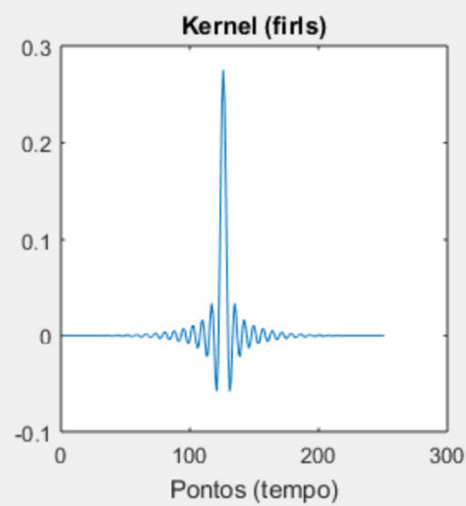
```
1 - clear
```

Figure 1

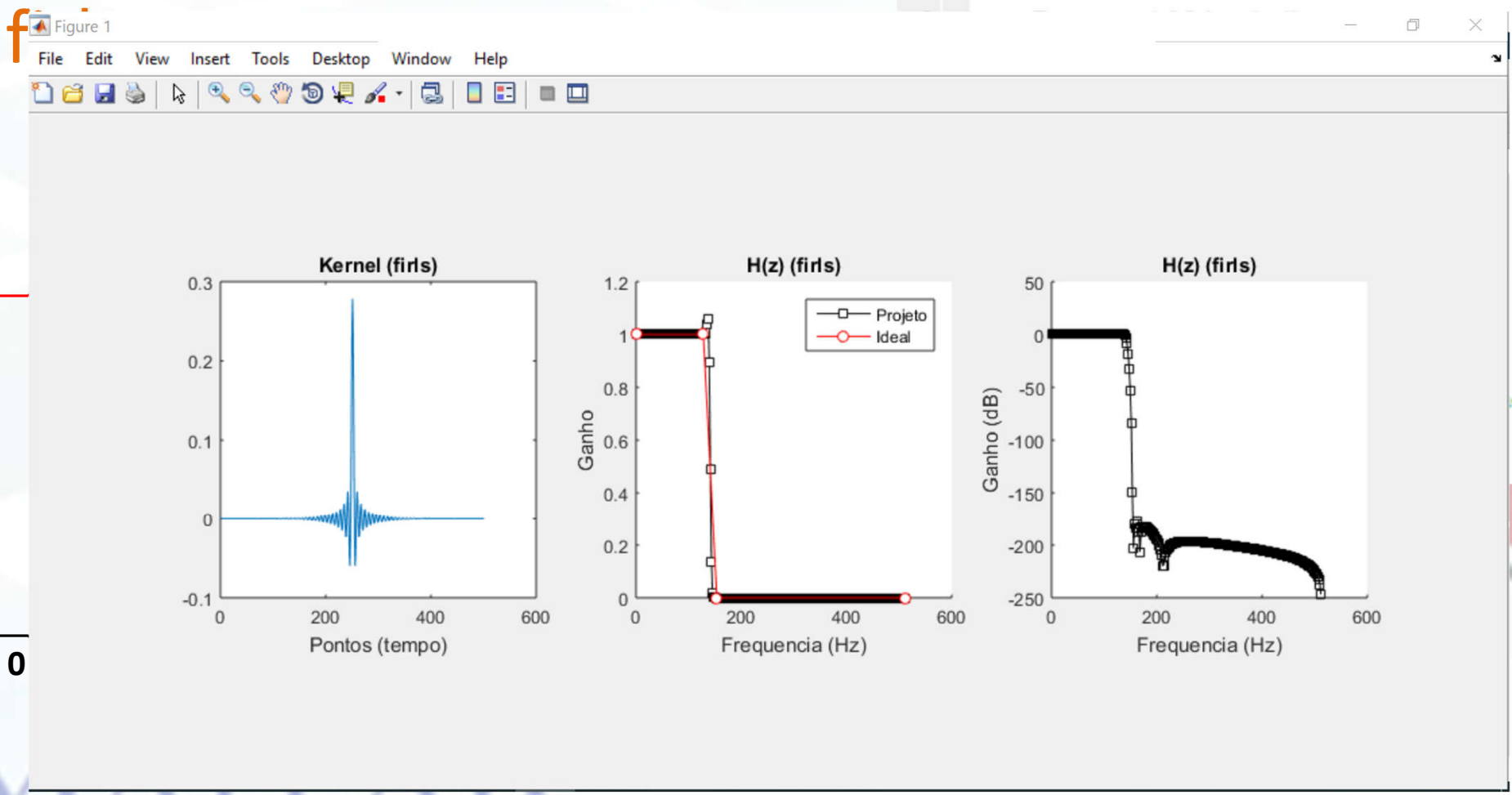
File Edit View Insert Tools Desktop Window Help



Ganho  
1  
0  
0



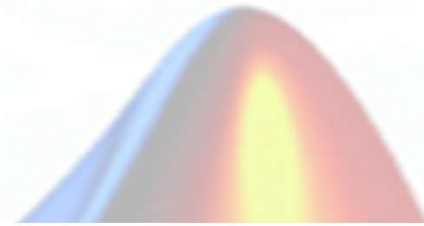
1 - clear



Ganho  
1  
0  
0

etria)

# fir1



## Description

`b = fir1(n,Wn)` uses a Hamming window to design an  $n$ th-order lowpass, bandpass, or multiband FIR filter with linear phase. The filter type depends on the number of elements of `Wn`. [example](#)

`b = fir1(n,Wn,ftype)` designs a lowpass, highpass, bandpass, bandstop, or multiband filter, depending on the value of `ftype` and the number of elements of `Wn`. [example](#)

### ▼ FIR Highpass Filter

Load `chirp.mat`. The file contains a signal, `y`, that has most of its power above  $F_s/4$ , or half the Nyquist frequency. The sample rate is 8192 Hz.

[View MATLAB Command](#)

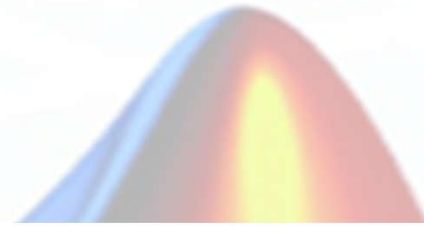
Design a 34th-order FIR highpass filter to attenuate the components of the signal below  $F_s/4$ . Use a cutoff frequency of 0.48 and a Chebyshev window with 30 dB of ripple.

```
load chirp

t = (0:length(y)-1)/Fs;

bhi = fir1(34,0.48,'high',chebwin(35,30));
freqz(bhi,1)
```

# fir1



## Description

`b = fir1(n,Wn)` uses a Hamming window to design an  $n$ th-order lowpass, bandpass, or multiband FIR filter with linear phase. The filter type depends on the number of elements of `Wn`.

[example](#)

`b = fir1(n,Wn,ftype)` designs a lowpass, highpass, bandpass, bandstop, or multiband FIR filter with linear phase.

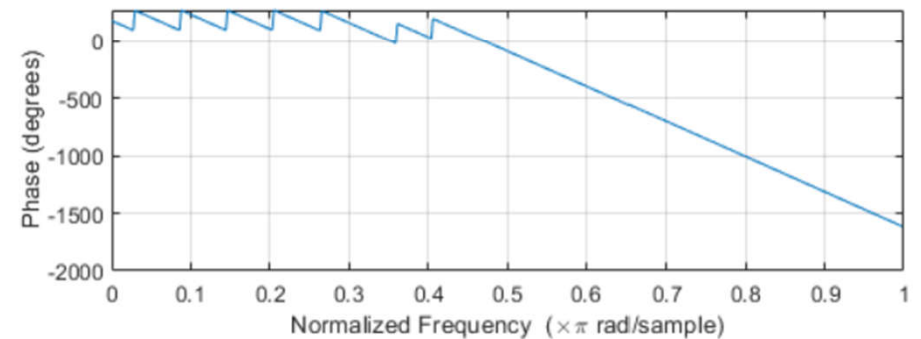
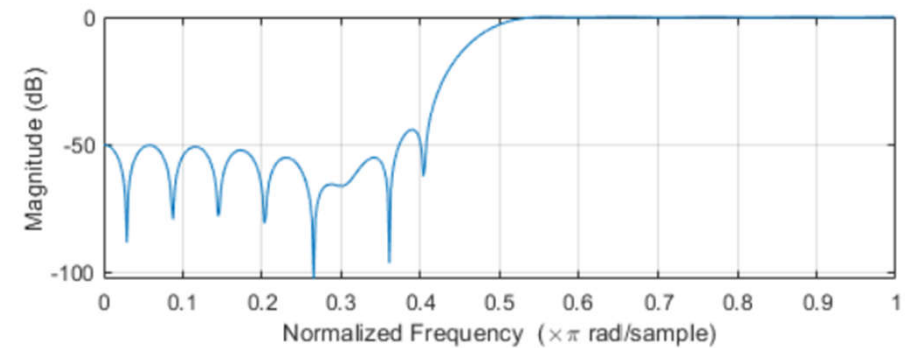
[example](#)

### ▼ FIR Highpass Filter

Load `chirp.mat`. The file contains a signal, `y`, that has most of its power above  $F_s$ /rate is 8192 Hz.

Design a 34th-order FIR highpass filter to attenuate the components of the signal below 0.48  $\pi$  rad/sample and a Chebyshev window with 30 dB of ripple.

```
load chirp
t = (0:length(y)-1)/Fs;
bhi = fir1(34,0.48,'high',chebwin(35,30));
freqz(bhi,1)
```



# fir1

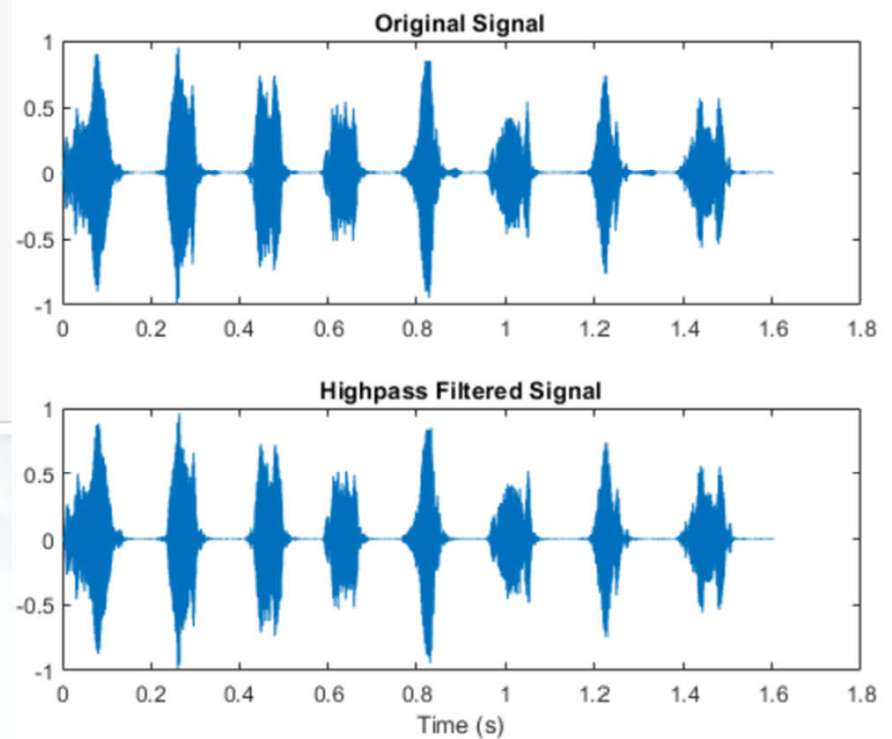


Filter the signal. Display the original and highpass-filtered signals. Use the same y-axis scale for both plots.

```
outhi = filter(bhi,1,y);

subplot(2,1,1)
plot(t,y)
title('Original Signal')
ys = ylim;

subplot(2,1,2)
plot(t,outhi)
title('Highpass Filtered Signal')
xlabel('Time (s)')
ylim(ys)
```



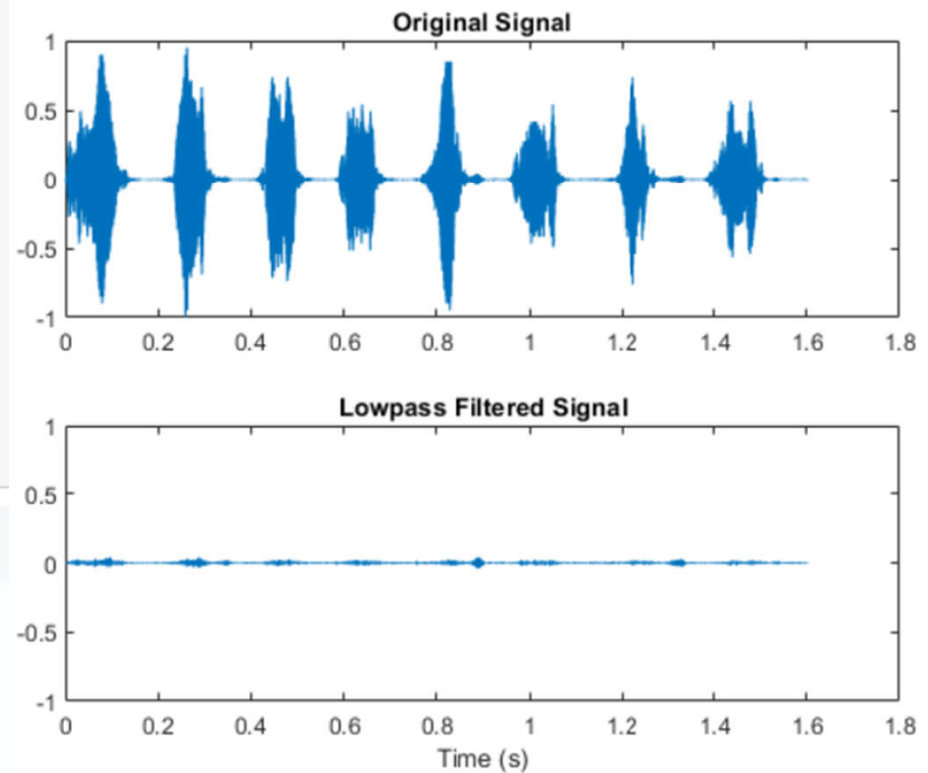
MATLAB®

# fir1



Design a lowpass filter with the same specifications. Filter the signal and compare the result to the original. Use the same y-axis scale for both plots.

```
blo = fir1(34,0.48,chebwin(35,30));  
  
outlo = filter(blo,1,y);  
  
subplot(2,1,1)  
plot(t,y)  
title('Original Signal')  
ys = ylim;  
  
subplot(2,1,2)  
plot(t,outlo)  
title('Lowpass Filtered Signal')  
xlabel('Time (s)')  
ylim(ys)
```



MATLAB®

Filtros digitais

# **DESIGN DE FILTROS DIGITAIS**

## **EXEMPLO IIR**

# butter

## Description

`[b,a] = butter(n,Wn)` returns the transfer function coefficients of an  $n$ th-order lowpass digital Butterworth filter with normalized cutoff frequency  $Wn$ .

[example](#)

`[b,a] = butter(n,Wn,fstype)` designs a lowpass, highpass, bandpass, or bandstop Butterworth filter, depending on the value of `fstype` and the number of elements of `Wn`. The resulting bandpass and bandstop designs are of order  $2n$ .

[example](#)

**Note:** See [Limitations](#) for information about numerical issues that affect forming the transfer function.

`[z,p,k] = butter(__)` designs a lowpass, highpass, bandpass, or bandstop digital Butterworth filter and returns its zeros, poles, and gain. This syntax can include any of the input arguments in previous syntaxes.

[example](#)

`[A,B,C,D] = butter(__)` designs a lowpass, highpass, bandpass, or bandstop digital Butterworth filter and returns the matrices that specify its state-space representation.

[example](#)

`[__] = butter(__,'s')` designs a lowpass, highpass, bandpass, or bandstop analog Butterworth filter with cutoff angular frequency  $Wn$ .

[example](#)

## Examples

[collapse all](#)

MATLAB®



# butter

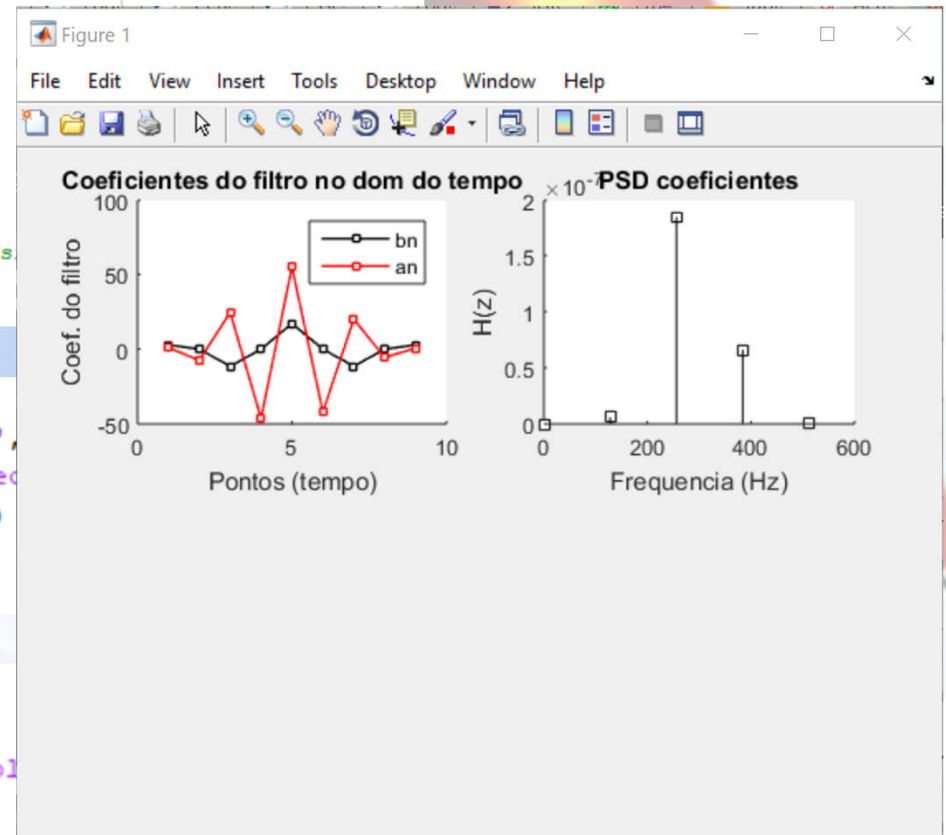
```
1 - clear
2 - Fs = 1024; % Hz
3 - nyquist = Fs/2;
4 - f = [20 45]/nyquist;
5 - ordem=4;
6
7 - % coeficientes do filtro (baseados na equação de diferenças)
8 - [bn,an] = butter(ordem,f);
9
10 - % H(z)- PSD do kernel
11 - psd_bn = abs(fft(bn)).^2;
12 - % Calcula o vetor de frequencias e exibe apenas a parte positiva (simetria)
13 - Hz = linspace(0,Fs/2,floor(length(bn)/2)+1);
14 - psd_bn = psd_bn(1:length(Hz));
```

```
16 - figure(1);
17 - subplot(2,2,1);hold on; plot(bn*1e5, 'ks-', 'linewidth',1, 'markersize',4, 'markerfacecolor', 'w');
18 - plot(an, 'rs-', 'linewidth',1, 'markersize',4, 'markerfacecolor', 'w')
19 - xlabel('Pontos (tempo)');ylabel('Coef. do filtro')
20 - title('Coeficientes do filtro no dom do tempo')
21 - legend({'bn'; 'an'})
```

```
23 - % H(z) (PSD)
24 - subplot(2,2,2), hold on
25 - stem(Hz,psd_bn(1:length(Hz)), 'ks-', 'markerfacecolor', 'w')
26 - xlabel('Frequencia (Hz)'), ylabel('H(z)')
27 - title('PSD coeficientes')
```

# butter

```
1 - clear
2 - Fs = 1024; % Hz
3 - nyquist = Fs/2;
4 - f = [20 45]/nyquist;
5 - ordem=4;
6
7 - % coeficientes do filtro (baseados na equação de diferenças)
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10 - % H(z)- PSD do kernel
11 - psd_bn = abs(fft(bn)).^2;
12 - % Calcula o vetor de frequencias e exibe apenas a parte positiva (s
13 - Hz = linspace(0,Fs/2,floor(length(bn)/2)+1);
14 - psd_bn = psd_bn(1:length(Hz));
15
16 - figure(1);
17 - subplot(2,2,1);hold on; plot(bn*1e5, 'ks-', 'linewidth',
18 - plot(an, 'rs-', 'linewidth', 1, 'markersize', 4, 'markerfacecol
19 - xlabel('Pontos (tempo)');ylabel('Coef. do filtro')
20 - title('Coeficientes do filtro no dom do tempo')
21 - legend({'bn'; 'an'})
22
23 - % H(z) (PSD)
24 - subplot(2,2,2), hold on
25 - stem(Hz,psd_bn(1:length(Hz)),'ks-', 'markerfacecol
26 - xlabel('Frequencia (Hz)'), ylabel('H(z)')
27 - title('PSD coeficientes')
```



```

29 % Obter a RI
30 - impulso = impseq(501,0,1000);
31 - hn=filter(bn,an,impulso);
32
33 %obter H(z) - PSD de hn
34 - psd_imp = abs(fft(hn)).^2;
35 - Hz = linspace(0,nyquist,floor(length(impulso)/2)+1);
36
37 - subplot(2,2,2), cla, hold on
38 - plot(impulso,'b','linewidth',1)
39 - plot(hn,'m','linewidth',1)
40 - set(gca,'xlim',[1 length(impulso)],'ylim',[-1 1]*.06)
41 - legend({'Impulso';'h[n]'})
42 - xlabel('Pontos (tempo)')
43 - title('RI')
44
45 % H(z)
46 - subplot(2,2,3), hold on
47 - plot(Hz,psd_imp(1:length(Hz)),'ms-','linewidth',1,'markerfacecolor','w','markersize',4)
48 - plot([0 f(1) f(1) f(2) f(2) 1]*nyquist,[0 0 1 1 0 0],'b','linewidth',2)
49 - set(gca,'xlim',[0 100])
50 - xlabel('Frequencia (Hz)'), ylabel('Atenuação')
51 - title('RF (Butterworth)')

```

```

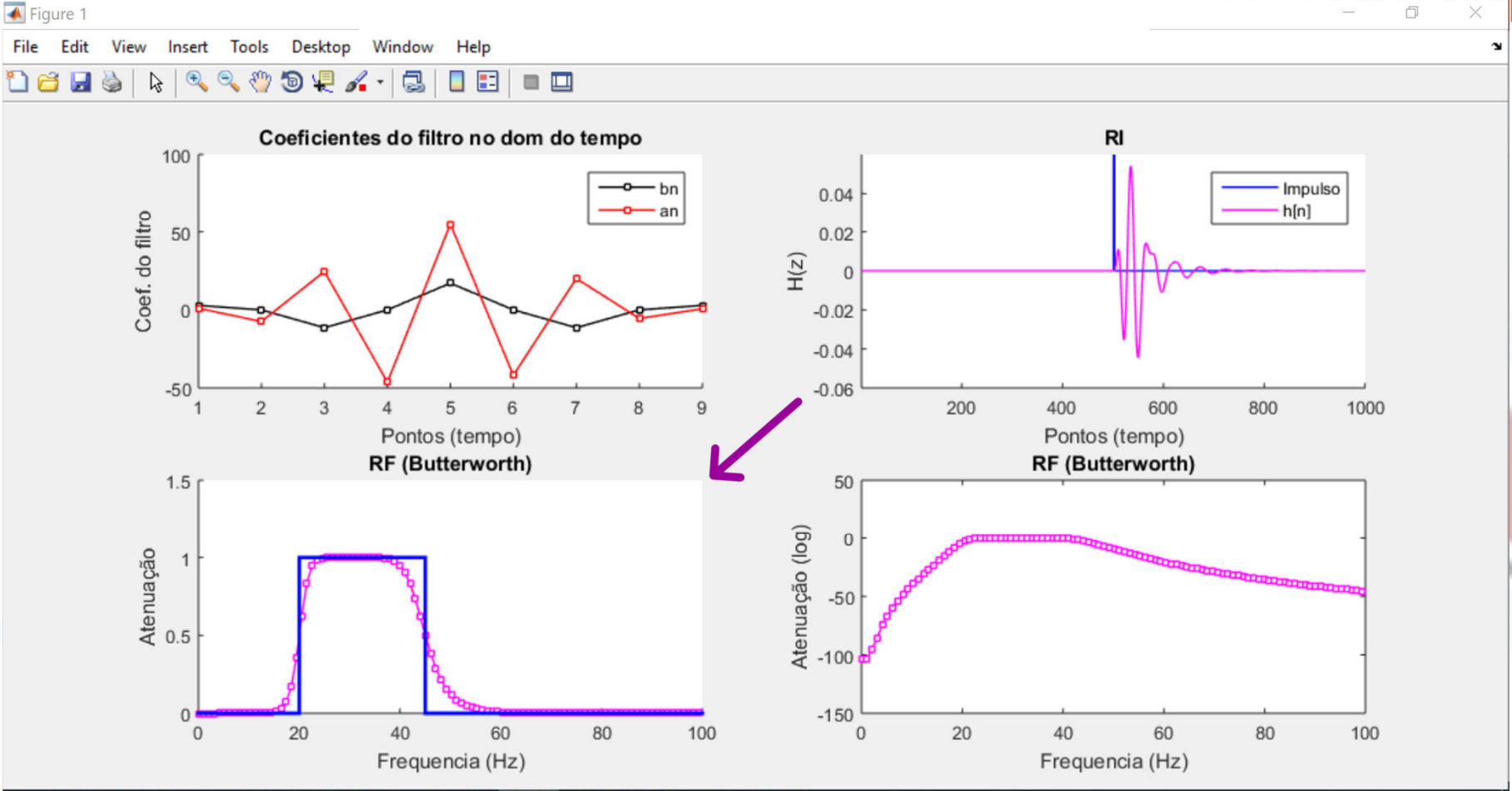
53 % H(z) em escala logaritmica
54 - subplot(2,2,4)
55 - plot(Hz,10*log10(psd_imp(1:length(Hz))),'ms-','linewidth',1,'markerfacecolor','w','markersize',4)
56 - set(gca,'xlim',[0 100])
57 - xlabel('Frequencia (Hz)'), ylabel('Atenuação (log)')
58 - title('RF (Butterworth)')

```



butter

```
29 % Obter a RI
30 impulso = impseq(501,0,1000);
31 hn=filter (bn,an,impulso);
```

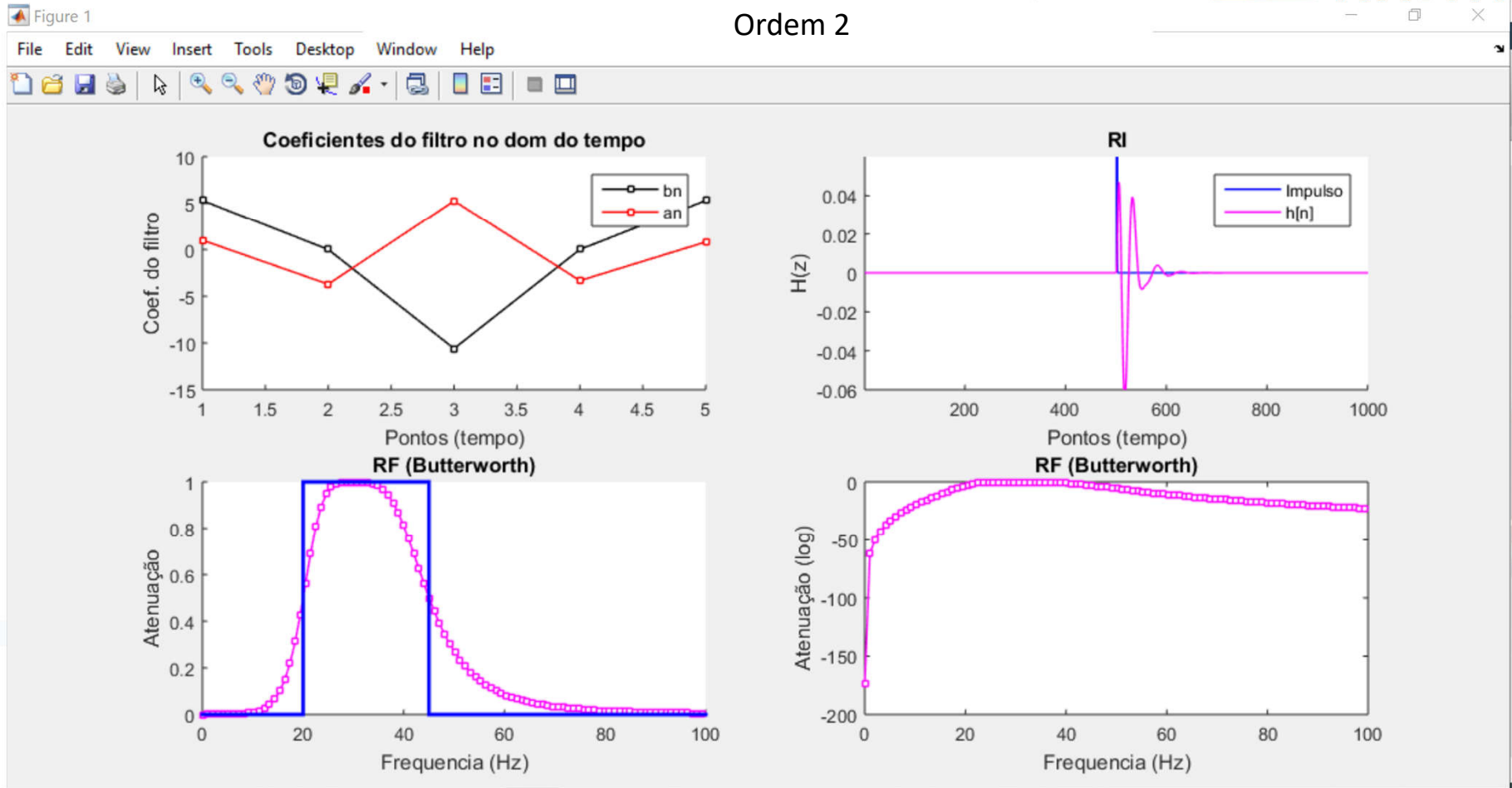


```
57 xlabel('Frequencia (Hz)'), ylabel('Atenuação (log)')
58 title('RF (Butterworth)')
```

```
29 % Obter a RI
30 impulso = impseq(501,0,1000);
31 hn=filter (bn,an,impulso);
```



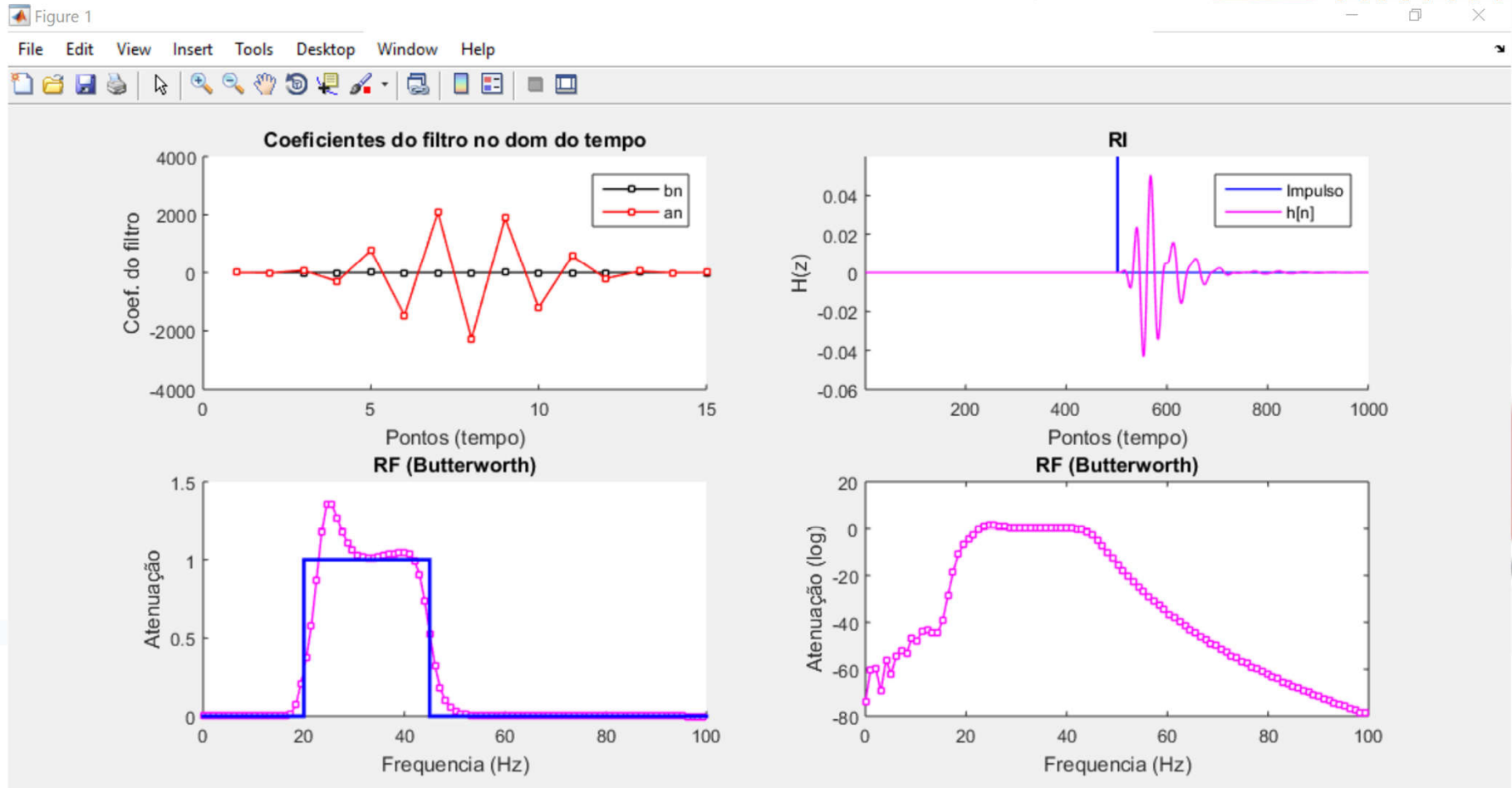
Ordem 2



```
57 xlabel('Frequencia (Hz)'), ylabel('Atenuação (log)')
58 title('RF (Butterworth)')
```

```
29 % Obter a RI
30 impulso = impseq(501,0,1000);
31 hn=filter (bn,an,impulso);
32
```

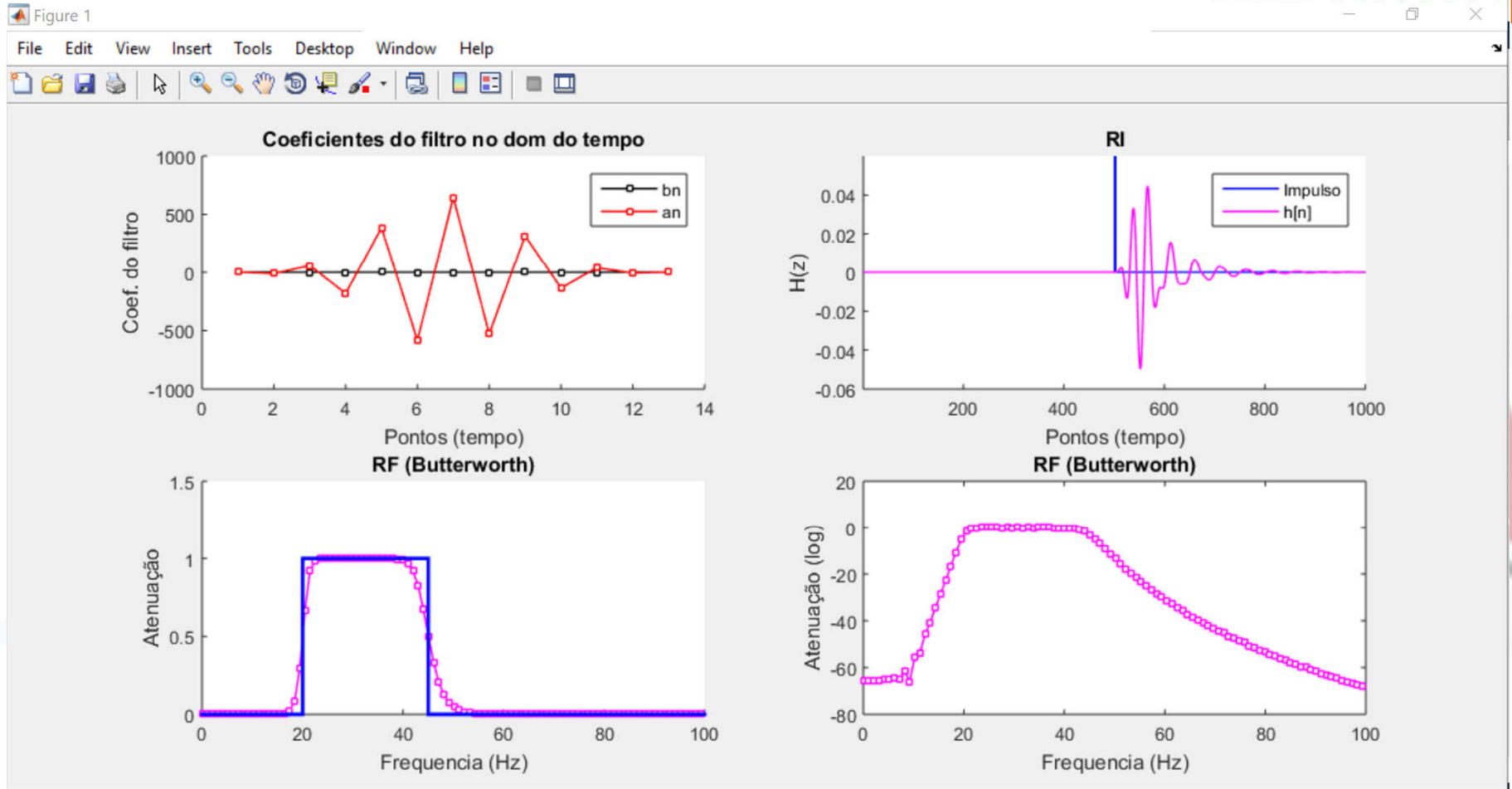
Ordem 7



```
53 xlabel('Frequencia (Hz)'), ylabel('Atenuação (log)')
54
55 title('RF (Butterworth)')
56
```

```
29 % Obter a RI
30 impulso = impseq(501,0,1000);
31 hn=filter (bn,an,impulso);
```

Ordem 6



```
57 xlabel('Frequencia (Hz)'), ylabel('Atenuação (log)')
58 title('RF (Butterworth)')
```