Fourier Transform: part 2 SCC0251/5830 – Image Processing

#### Prof. Moacir Ponti

#### Instituto de Ciências Matemáticas e de Computação - USP

2021/1





- Discrete Fourier Transform
  - Convolution Theorem
- 3 Filtering in frequency domain
- 4 Fast Fourier Transform (FFT)

Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

< ∃ >

- ∢ /⊐ ►

Sac

Moacir P

It allows us to see the frequency content of a given signal.

$$F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j2\pi t\omega} dt$$
  
= 
$$\int_{-\infty}^{\infty} f(t) \cos(2\pi t\omega) dt + j \int_{-\infty}^{\infty} f(t) \sin(2\pi t\omega) dt$$
  
$$\int_{-\frac{1}{2}}^{\frac{1}{2}} \frac{1}{\frac{1}{2}} \frac{1}{\frac{1}{2}}$$

It allows us to see the frequency content of a given signal.





æ

590

5/43

Moacir Ponti (ICMC-USP)

Fourier Transform

 $\exists \mapsto$ 2021/1

(日)



Moacir Ponti (ICMC-USP)

#### Discrete Fourier Transform

for N samples:

$$F[k] = \sum_{n=0}^{N-1} f[n] e^{-j2\pi n \frac{k}{N}}$$
$$= \frac{A[k]}{A[k]} + j \frac{B[k]}{B[k]}$$

Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

æ

590

6/43

イロト イポト イヨト イヨト

For each frequency a complex exponential with:

- the relative amplitude of the cosine (real part) and of the sine (imaginary part) as a function of different frequencies,
- the representation of the signal in the **frequency domain**:
  - A[k] = Re(F[k])
  - B[k] = Im(F[k])



$$F[k] = \sum_{n=0}^{N-1} f[n] e^{-j2\pi n \frac{k}{N}}$$

• evaluating F[k] on different k, we obtain the **amplitudes of** cosines (real part) and sines (imaginary part) so that we can reconstruct f[n] if needed.

Moacir Ponti (ICMC–USP)

A B A B A
 A
 B
 A
 A
 B
 A
 A
 B
 A
 A
 B
 A
 A
 B
 A
 A
 B
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A
 A

8/43



$$|F[k]| = \sqrt{A[k]^2 + B[k]^2}$$

$$\phi(F[k]) = \arctan\left(\frac{A[k]}{B[k]}\right)$$

◆□▶ ◆□▶ ◆□▶ ◆□▶

Moacir Ponti (ICMC-USP)

2021/1

æ

୬ ଏ ୯ ୨ / 43

# Fourier Transform: practical considerations

- Cosine is an odd (anti-symmetric) function
- Sine is an even (symmetric) function

Because of that, values of k after N/2 start to repeat. That is why we just show half the values.



# Fourier Transform: practical considerations

- Cosine is an odd (anti-symmetric) function
- Sine is an even (symmetric) function

Because of that, values of k after N/2 start to repeat. That is why we just show half the values.



The maximum frequency N/2 has a important relationship with the Nyquist-Shannon Theorem of sampling.

#### **Big Picture**

# Fourier Transform: noise and abrupt signals

- Because we start to measure at k = 0 and stop at k = N 1, this can be considered high frequency content, hampering the analysis
- To prevent that, it is possible to use a windowing function, to obtain a signal that increases its amplitude more slowly



# Fourier Transform: noise and abrupt signals



Moacir Ponti (ICMC–USP)

2021/1

୬ ୯.୯ 12 / 43





- 2 Discrete Fourier TransformConvolution Theorem
  - 3 Filtering in frequency domain
- 4 Fast Fourier Transform (FFT)

Moacir Ponti (ICMC-USP)

Fourier Transform

E ► < E ► 4</p>
2021/1

Image: A match a ma

Sac

13/43

э

# Discrete Fourier Transform 2D (DFT)

$$F(u,v) = \frac{1}{\sqrt{mn}} \sum_{x=0}^{n-1} \sum_{y=0}^{m-1} f(x,y) e^{-j2\pi(ux/n+vy/m)},$$

f(x, y) is a function representing an image with size  $n \times m$ .

イロトイラトイミト モン つくつ Fourier Transform 2021/1 14/43

Moacir Ponti (ICMC–USP)

# Discrete Fourier Transform 2D (DFT)

$$F(u,v) = \frac{1}{\sqrt{mn}} \sum_{x=0}^{n-1} \sum_{y=0}^{m-1} f(x,y) e^{-j2\pi(ux/n+vy/m)},$$

f(x, y) is a function representing an image with size  $n \times m$ .

• considering the 2-D case: x, y are coordinates, u, v are frequencies in each direction.

Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

# Discrete Fourier Transform 2D (DFT)

$$F(u,v) = \frac{1}{\sqrt{mn}} \sum_{x=0}^{n-1} \sum_{y=0}^{m-1} f(x,y) e^{-j2\pi(ux/n+vy/m)},$$

f(x, y) is a function representing an image with size  $n \times m$ .

- considering the 2-D case: x, y are coordinates, u, v are frequencies in each direction.
- as in 1-D, we evaluate it for a range of *u* and *v*:
  - u = 0, 1, ..., n 1 and v = 0, 1, ..., m 1

2021/1

# Inverse Discrete Fourier Transform 2D (IDFT)

$$f(x,y) = \frac{1}{\sqrt{nm}} \sum_{u=0}^{n-1} \sum_{v=0}^{m-1} F(u,v) e^{j2\pi(ux/n+vy/m)},$$
  
for  $x = 0, 1, ..., n-1$  and  $y = 0, 1, ..., m-1$ 

Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1 15/43

イロト イポト イヨト イヨト

э

Sac

• Writing the DFT-2D in the polar form

$$F(u,v) = |F(u,v)|e^{j\phi(u,v)},$$



э

Sac

16/43

< ≣⇒

2021/1

Image: A match a ma

Moacir Ponti (ICMC–USP)

• Writing the DFT-2D in the polar form

$$F(u,v) = |F(u,v)|e^{j\phi(u,v)},$$

• the magnitude is often called *Fourier spectrum* (or frequency spectrum):

$$|F(u, v)| = [R^2(u, v) + I^2(u, v)]^{1/2}$$

Moacir Ponti (ICMC–USP)

(B)

2021/1

16/43

Image: Image:

• Writing the DFT-2D in the polar form

$$F(u,v) = |F(u,v)|e^{j\phi(u,v)},$$

• the magnitude is often called *Fourier spectrum* (or frequency spectrum):

$$|F(u, v)| = [R^{2}(u, v) + I^{2}(u, v)]^{1/2}$$

• the *phase angle* is:

$$\phi(u, v) = \arctan\left[\frac{l(u, v)}{R(u, v)}\right]$$

Moacir Ponti (ICMC-USP)

16 / 43

< ∃ >

• Writing the DFT-2D in the polar form

$$F(u,v) = |F(u,v)|e^{j\phi(u,v)},$$

• the magnitude is often called *Fourier spectrum* (or frequency spectrum):

$$|F(u, v)| = [R^{2}(u, v) + I^{2}(u, v)]^{1/2}$$

• the *phase angle* is:

$$\phi(u, v) = \arctan\left[\frac{I(u, v)}{R(u, v)}\right]$$

• the *power spectrum* is:

$$P(u,v) = |F(u,v)|^2$$

Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

#### Properties

• From Fourier Transform

$$F(0,0) = \frac{1}{\sqrt{nm}} \sum_{x=0}^{n-1} \sum_{y=0}^{m-1} f(x,y),$$

the frequency term zero is proportional to the sum of all values of f(x, y) (normalised by  $1/\sqrt{nm}$ ).

3

17 / 43

イロト イポト イヨト イヨト

#### Properties

• From Fourier Transform

$$F(0,0) = \frac{1}{\sqrt{nm}} \sum_{x=0}^{n-1} \sum_{y=0}^{m-1} f(x,y),$$

the frequency term zero is proportional to the sum of all values of f(x, y) (normalised by  $1/\sqrt{nm}$ ).

• The Fourier spectrum of a real function is even (symmetric with respect to the origin):

$$|F(u,v)| = |F(-u,-v)|$$

Moacir Ponti (ICMC–USP)

17 / 43

#### Properties

• From Fourier Transform

$$F(0,0) = \frac{1}{\sqrt{nm}} \sum_{x=0}^{n-1} \sum_{y=0}^{m-1} f(x,y),$$

the frequency term zero is proportional to the sum of all values of f(x, y) (normalised by  $1/\sqrt{nm}$ ).

• The Fourier spectrum of a real function is even (symmetric with respect to the origin):

$$|F(u,v)| = |F(-u,-v)|$$

• the phase angle is odd (anti-symmetric):

Moacir Ponti (ICMC–USP)

2021/1

17/43

Plotting the Fourier spectrum

• Re-positioning of the quadrants (FFT shift) by using coordinates  $(-1)^{x+y}$ 



Shift(|F(u, v)|)



Fourier Transform

- 4 回 ト 4 ヨト 2021/1

Plotting the Fourier spectrum

- Re-positioning of the quadrants (FFT shift) by using coordinates  $(-1)^{x+y}$
- Applying the logarithm:  $(\log(1 + |F(u, v)|))$  to increase the dynamic range of frequencies with small amplitude.







æ

Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

◆□▶ ◆□▶ ◆□▶ ◆□▶

୬ ୯ ୯ 19 / 43



Moacir Ponti (ICMC-USP)

Fourier Transform



Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

◆□▶ ◆□▶ ◆□▶ ◆□▶

୬ ୯.୯ 19 / 43

E





2021/1

◆□▶ ◆□▶ ◆□▶ ◆□▶

୬ ୯.୯ 19 / 43

1

# Visualizing the Fourier Spectrum

After FFT-Shift:

- In the central point of the image we have lower frequencies (with respect to u, v)
- Higher frequencies are further from the centre



Moacir Ponti (ICMC-USP)

Fourier Transform

2021/1





- 2 Discrete Fourier Transform
  - Convolution Theorem
- 3 Filtering in frequency domain



Sac

#### Convolution Theorem

• A circular convolution in 2D is given by:

$$f(x,y) * h(x,y) = \sum_{i=0}^{n-1} \sum_{j=0}^{m-1} f(i,j)h(x-i,y-j),$$

Moacir Ponti (ICMC–USP)

э

イロト イポト イヨト イヨト

2021/1

Sac

#### Convolution Theorem

• The 2D convolution theorem has the following result :

$$f(x,y) * h(x,y) \Leftrightarrow F(u,v) \cdot H(u,v),$$

$$f(x,y) \cdot h(x,y) \Leftrightarrow F(u,v) * H(u,v).$$

F and H are the Fourier Transforms of f and h, respectively

Image: A matrix

23 / 43

- ( E
## Agenda



- Discrete Fourier TransformConvolution Theorem
- 3 Filtering in frequency domain



Moacir Ponti (ICMC-USP)

Fourier Transform

2021/1

イロト イボト イヨト イヨト

Sac

24 / 43

э

• Modify the Fourier transform of an image and calculate its inverse:

$$g(x,y) = \mathfrak{F}^{-1}\left[F(u,v)H(u,v)\right]$$



-

2021/1

Sac

25 / 43

Moacir Ponti (ICMC–USP)

Fourier Transform

**1** Input: f(x, y) with size  $m \times n$ 



Sac

26/43

Moacir Ponti (ICMC-USP)

Fourier Transform

2021/1

→ Ξ →

Image: A match a ma

- 1 Input: f(x, y) with size  $m \times n$
- **2** Obtain optimal values to perform the transform: P and Q. Options:
  - use the next power of two, or
  - use P = 2n and Q = 2m.

- 1 Input: f(x, y) with size  $m \times n$
- **2** Obtain optimal values to perform the transform: P and Q. Options:
  - use the next power of two, or
  - use P = 2n and Q = 2m.
- Solution Create an image  $f_{\rho}(x, y)$  of size  $P \times Q$ , padding with zeros,

- 1 Input: f(x, y) with size  $m \times n$
- **2** Obtain optimal values to perform the transform: P and Q. Options:
  - use the next power of two, or
  - use P = 2n and Q = 2m.
- Solution Create an image  $f_p(x, y)$  of size  $P \times Q$ , padding with zeros,
- Compute DFT  $F(u, v) = \mathfrak{F}[f_p(x, y)]$ ,

- 1 Input: f(x, y) with size  $m \times n$
- **2** Obtain optimal values to perform the transform: P and Q. Options:
  - use the next power of two, or
  - use P = 2n and Q = 2m.
- Solution Create an image  $f_p(x, y)$  of size  $P \times Q$ , padding with zeros,
- Compute DFT  $F(u, v) = \mathfrak{F}[f_p(x, y)],$
- Use a symmetric filter H(u, v) of size  $P \times Q$  centred in (P/2, Q/2),

- 1 Input: f(x, y) with size  $m \times n$
- **2** Obtain optimal values to perform the transform: P and Q. Options:
  - use the next power of two, or
  - use P = 2n and Q = 2m.
- Solution Create an image  $f_p(x, y)$  of size  $P \times Q$ , padding with zeros,
- Compute DFT  $F(u, v) = \mathfrak{F}[f_p(x, y)],$
- Use a symmetric filter H(u, v) of size  $P \times Q$  centred in (P/2, Q/2),
- Compute processed image:  $g_p(x, y) = real(\mathfrak{F}^{-1}[F(u, v)H(u, v)])$ ,

2021/1

- 1 Input: f(x, y) with size  $m \times n$
- **2** Obtain optimal values to perform the transform: P and Q. Options:
  - use the next power of two, or
  - use P = 2n and Q = 2m.
- Solution Create an image  $f_p(x, y)$  of size  $P \times Q$ , padding with zeros,
- Compute DFT  $F(u, v) = \mathfrak{F}[f_p(x, y)],$
- Use a symmetric filter H(u, v) of size  $P \times Q$  centred in (P/2, Q/2),
- Compute processed image:  $g_p(x, y) = real(\mathfrak{F}^{-1}[F(u, v)H(u, v)])$ ,
- **Output:** g(x, y) size  $n \times m$ , after removing the zero-padded region.

26 / 43

< 日 > < 同 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < 回 > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ >

- (Low-Pass filter):
  - Allow the low frequencies to "pass", retaining higher frequencies
  - An "ideal" filter removes all frequencies beyond a given threshold



- (*High-Pass filter*):
  - remove lower frequencies, while allowing to "pass" higher frequencies
  - An "ideal" high-pass filter removes all frequencies below a given threshold



• (Bandpass filter):

• select a range of frequencies to be kept



• (Band-stop filter):

• select a range of frequencies to be removed



Sac

# High-pass filter





2021/1

3

< ∃ >

< 47 ▶

990

#### Low-pass filter





2021/1

Sac 32 / 43

### Band-stop filter



Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

E

SQC

### Band-stop filter



Moacir Ponti (ICMC-USP)

Fourier Transform

# Band-stop filter



### Agenda

#### 1 Big Picture

- Discrete Fourier TransformConvolution Theorem
- 3 Filtering in frequency domain



Moacir Ponti (ICMC–USP)

Fourier Transform

2021/1

イロト イボト イヨト イヨト

୬ ୯ ୯ 34 / 43

э

## Fast Fourier Transform (FFT)

• Efficient algorithm to compute DFT



イロト イポト イヨト イヨト

2021/1

Sac

35 / 43

Moacir Ponti (ICMC-USP)

Fourier Transform

## Fast Fourier Transform (FFT)

- Efficient algorithm to compute DFT
- Uses ideas from Gauss (1805)

→

Image: A match a ma

3

- Efficient algorithm to compute DFT
- Uses ideas from Gauss (1805)
- The FFT Algorithm was published by Cooley and Tukey, 1965.





- James W. Cooley (1926–2016):
  - BSc., MSc., PhD. Mathematics;
  - Computer programmer with von Neumann at U.Princeton (1953-1956);
  - Researcher at IBM until 1991, his last position was professor at University of Rhode Island.
- John Tukey (1915-2000):
  - BSc., MSc. in Chemistry; PhD. Mathematics;
  - Professor at U.Princeton;
  - Researcher at AT&T Bell Labs;
  - Also known for developing (*boxplot*) and the *post-hoc* Tukey HSD test.

2021/1

• Allows to compute Fourier Transform in  $O(N \log_2 N)$ 



イロト イポト イヨト イヨト

2021/1

Sac

37 / 43

Moacir Ponti (ICMC-USP)

Fourier Transform

- Allows to compute Fourier Transform in  $O(N \log_2 N)$
- There are several ways to implement. Originally assumed sample size in powers of two.



- Allows to compute Fourier Transform in  $O(N \log_2 N)$
- There are several ways to implement. Originally assumed sample size in powers of two.
- The original algorithm uses the *successive duplication method*.
  - note that the DFT is a sequence of products and sums:

$$F(u) = \sum_{x=0}^{N-1} f(x) e^{-j\frac{2\pi u}{N}x},$$

Moacir Ponti (ICMC–USP)

37 / 43

- Allows to compute Fourier Transform in  $O(N \log_2 N)$
- There are several ways to implement. Originally assumed sample size in powers of two.
- The original algorithm uses the *successive duplication method*.
  - note that the DFT is a sequence of products and sums:

$$F(u) = \sum_{x=0}^{N-1} f(x) e^{-j\frac{2\pi u}{N}x},$$

2021/1

37 / 43

• however, the term  $e^{-j2\pi}$  is always the same, with different powers of (u/N)x

Moacir Ponti (ICMC–USP)

• Re-writing:

$$F(u) = \sum_{x=0}^{N-1} f(x) W_N^{ux},$$

where 
$$W_N = e^{-jrac{2\pi}{N}}$$

Properties:

æ

590

38 / 43

◆□▶ ◆□▶ ◆□▶ ◆□▶

2021/1

Moacir Ponti (ICMC–USP)

Fourier Transform

• Re-writing:

$$F(u) = \sum_{x=0}^{N-1} f(x) W_N^{ux},$$

where 
$$W_{N}=e^{-jrac{2\pi}{N}}$$
 ,

Properties:

• Complex conjugate symmetry:

$$W_N^{u(N-x)} = W_N^{-ux} = (W_N^{ux})^*$$
(1)

イロト イポト イヨト イヨト

2021/1

Moacir Ponti (ICMC-USP)

3 38 / 43

Sac

• Re-writing:

$$F(u) = \sum_{x=0}^{N-1} f(x) W_N^{ux},$$

where 
$$W_N = e^{-j\frac{2\pi}{N}}$$
, and  $W_N^{uN} = e^{-j2\pi u} = 1$ 

Properties:

• Complex conjugate symmetry:

$$W_N^{u(N-x)} = W_N^{-ux} = (W_N^{ux})^*$$
(1)

イロト イポト イヨト イヨト

2021/1

Sac

38 / 43

3

Moacir Ponti (ICMC–USP)

• Re-writing:

$$F(u) = \sum_{x=0}^{N-1} f(x) W_N^{ux},$$

where 
$$W_N = e^{-j\frac{2\pi}{N}}$$
, and  $W_N^{uN} = e^{-j2\pi u} = 1$ 

Properties:

• Complex conjugate symmetry:

$$W_N^{u(N-x)} = W_N^{-ux} = (W_N^{ux})^*$$
 (1)

- 4 🗗 ▶

**Periodicity** in *x*, *u*:

$$W_N^{ux} = W_N^{u(x+N)} = W_N^{(u+N)x}$$

Moacir Ponti (ICMC–USP)

э

< ∃ >

2

590

• Decimate: obtain DFT using smaller parts assuming  $N = 2^m$ 

decompose f(x) into indexes:
even (index x = 2r) and odd (index x = 2r + 1):

$$F(u) = \sum_{r=0}^{(N/2)-1} f(2r) W_N^{u2r} + \sum_{r=0}^{(N/2)-1} f(2r+1) W_N^{u(2r+1)},$$

Moacir Ponti (ICMC–USP)

2021/1 39/43

3

《曰》 《圖》 《臣》 《臣》

• Decimate: obtain DFT using smaller parts assuming  $N = 2^m$ 

decompose f(x) into indexes:
even (index x = 2r) and odd (index x = 2r + 1):

$$F(u) = \sum_{r=0}^{(N/2)-1} f(2r) W_N^{u2r} + \sum_{r=0}^{(N/2)-1} f(2r+1) W_N^{u(2r+1)},$$
  
= 
$$\sum_{r=0}^{(N/2)-1} f(2r) W_N^{u2r} + W_N^u \sum_{r=0}^{(N/2)-1} f(2r+1) W_N^{u2r},$$

《曰》 《圖》 《臣》 《臣》

• Decimate: obtain DFT using smaller parts assuming  $N = 2^m$ 

decompose f(x) into indexes:
 even (index x = 2r) and odd (index x = 2r + 1):

$$F(u) = \sum_{r=0}^{(N/2)-1} f(2r) W_N^{u2r} + \sum_{r=0}^{(N/2)-1} f(2r+1) W_N^{u(2r+1)},$$
  
$$= \sum_{r=0}^{(N/2)-1} f(2r) W_N^{u2r} + W_N^u \sum_{r=0}^{(N/2)-1} f(2r+1) W_N^{u2r},$$
  
$$= \sum_{r=0}^{(N/2)-1} f(2r) (W_N^2)^{ux} + W_N^u \sum_{r=0}^{(N/2)-1} f(2r+1) (W_N^2)^{ux},$$

Moacir Ponti (ICMC–USP)

2021/1

3

39 / 43

《曰》 《圖》 《臣》 《臣》

• Due to the periodicity of DFT:

$$F\left(u+\frac{N}{2}\right)=F(u)$$

• We can write the transform as:

$$F(u) = \begin{cases} F_e(u) + W_N^{ux} F_o(u), & \text{for } 0 \le u < N/2 \\ F_e(u - N/2) + W_N^{ux} F_o(u - N/2), & \text{for } N/2 \le u < N/2 \end{cases}$$

Moacir Ponti (ICMC–USP)

40 / 43

< E.

2021/1

Image: Image:

Sac

• By using the relationship between the terms W:

$$W_N^2 = e^{-j\frac{2\pi}{N}^2} = e^{-j\frac{4\pi}{N}} = e^{-j\frac{2\pi}{N/2}} = W_{N/2}$$

• Split the DFT into a sum of 2 DFTs with even and odd indices:

$$F(u) = \sum_{r=0}^{(N/2)-1} f(2r) W_{N/2}^{ux} + W_N^u \sum_{r=0}^{(N/2)-1} f(2r+1) W_{N/2}^{ux},$$
  
$$F(u) = F_e(u) + W_N^u \cdot F_o(u)$$

Moacir Ponti (ICMC–USP)

41 / 43

Image: A mathematical states and a mathem
- For each N/2 we now have  $2 \cdot N/4$
- Repeating the procedure

$$1:\frac{N}{2} \to 2\left(\frac{N}{2}\right)^2 + N = \frac{N^2}{2} + N$$



Moacir Ponti (ICMC-USP)

→

- For each N/2 we now have  $2 \cdot N/4$
- Repeating the procedure

$$1: \frac{N}{2} \to 2\left(\frac{N}{2}\right)^2 + N = \frac{N^2}{2} + N$$
$$2: \frac{N}{4} \to 2\left(2\left(\frac{N}{4}\right)^2 + \frac{N}{2}\right)^2 + N = \frac{N^2}{4} + 2N$$

Moacir Ponti (ICMC-USP)

2021/1 42/43

Э

→

- For each N/2 we now have  $2 \cdot N/4$
- Repeating the procedure

$$1: \frac{N}{2} \to 2\left(\frac{N}{2}\right)^2 + N = \frac{N^2}{2} + N$$
$$2: \frac{N}{4} \to 2\left(2\left(\frac{N}{4}\right)^2 + \frac{N}{2}\right)^2 + N = \frac{N^2}{4} + 2N$$
$$3: \frac{N}{8} \to 2\left(2\left(2\left(\frac{N}{8}\right)^2 + \frac{N}{4}\right) + \frac{N}{2}\right) + N = \frac{N^2}{8} + 3N$$

→

2021/1

- For each N/2 we now have  $2 \cdot N/4$
- Repeating the procedure

. . . . . .

$$1: \frac{N}{2} \to 2\left(\frac{N}{2}\right)^2 + N = \frac{N^2}{2} + N$$
$$2: \frac{N}{4} \to 2\left(2\left(\frac{N}{4}\right)^2 + \frac{N}{2}\right)^2 + N = \frac{N^2}{4} + 2N$$
$$3: \frac{N}{8} \to 2\left(2\left(2\left(\frac{N}{8}\right)^2 + \frac{N}{4}\right) + \frac{N}{2}\right) + N = \frac{N^2}{8} + 3N$$

I ∃ ≥

2021/1

Image: A match a ma

- For each N/2 we now have  $2 \cdot N/4$
- Repeating the procedure

$$1: \frac{N}{2} \to 2\left(\frac{N}{2}\right)^2 + N = \frac{N^2}{2} + N$$
$$2: \frac{N}{4} \to 2\left(2\left(\frac{N}{4}\right)^2 + \frac{N}{2}\right)^2 + N = \frac{N^2}{4} + 2N$$
$$3: \frac{N}{8} \to 2\left(2\left(2\left(\frac{N}{8}\right)^2 + \frac{N}{4}\right) + \frac{N}{2}\right) + N = \frac{N^2}{8} + 3N$$

. . . . . .

$$P: rac{N}{2^P} 
ightarrow rac{N^2}{2^P} + pN = N + N \log_2 N$$

Moacir Ponti (ICMC–USP)

2021/1 42/43

3

イロト イポト イヨト イヨト

# Bibliography I



#### GONZALEZ, R.C.; WOODS, R.E. \*

## Processamento Digital de Imagens, 3.ed

Capítulo 4. Pearson, 2010.

Moacir Ponti (ICMC-USP)

Fourier Transform

Sac

43 / 43

< ∃ >

2021/1

< 4<sup>™</sup> > -